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A Biomedical Application by Using Optimal Fuzzy Sliding-Mode Control

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1. Introduction

The development of biochips is a major thrust of the rapidly growing biotechnology industry. Research on biomedical or biochemical analysis miniaturization and integration has made explosive progress by using biochips recently. For example, capillary electrophoresis (CE), sample preconcentration, genomic DNA extraction, and DNA hybridization have been successfully miniaturized and operated in a single-step chip. However, there is still a considerable technical challenge in integrating these procedures into a multiple-step system. In biometric and biomedical applications, the special transporting mechanism must be designed for the μ TAS (micro total analysis system) to move samples and reagents through the microchannels that connect the unit procedure components in the system. Therefore, an important issue for this miniaturization and integration is microfluid management technique, i.e., microfluid transportation, metering, and mixing. This chapter introduced a method to achieve the microfluidic manipulated implementation on biochip system with a pneumatic pumping actuator and a feedback-signal flowmeter by using an optimal fuzzy sliding-mode control (OFSMC) design based on the 8051 microprocessor.

However, the relationships of the pumping mechanisms, the operating conditions of the devices, and the transporting behavior of the multi-component fluids in these channels are quite complicated. Because the main disadvantages of the mechanical valves utilized moving parts are the complexity and expense of fabrication, and the fragility of the components. Therefore, a novel recursively-structured apparatus of valveless microfluid manipulating method based on a pneumatic pumping mechanism has been utilized in this study. The working principle of this pumping design on this device should not directly relate to the nature of the fluid components. The driving force acting on the microliquid drop in the microchannel of this device is based on the pneumatic pumping which is induced by a blowing airflow. Furthermore, the pneumatic pumping actuator should be independent of the actuation responsible for the biochemical analysis on the chip system, so the contamination of pneumatic pumping source can be avoided. The total biochip mechanism consists of an external pneumatic actuator and an on-chip planar structure for airflow reception.

In order to achieve microfluidic manipulation in the microchannel of the biochip system, pneumatic pumping controller plays an important role. Therefore, a design of the controller

has been investigated numerically and experimentally in the present charter. In the control structure of biochip system, at first, the mathematical model of the biochip mechanism is identified by ARX model. Second, according to the results of the biochip-mechanism identification, the control-algorithm design is developed. By the simulation results of the biochip system with a feedback-signals flowmeter, they show the effectiveness of the developed control algorithm. Third, architecture of the control algorithm is integrated on a microprocessor to implement microfluidic manipulation. Since the mathematical model of the flow control mechanism in the biochip microchannels is a complicated nonlinear plant, the fuzzy logic control (FLC) design of the controller will be utilized. Design of the FLC based on the fuzzy set theory has been widely applied to consumer products or industrial process controls. In particular, they are very effective techniques for complicated, nonlinear, and imprecise plants for which either no mathematical model exists or the mathematical model is severely nonlinear. The FLC can approximate the human expert's control behaviors to work fine in such ill-defined environments. For some applications, the FLC can be divided into two classes 1) the general-purpose fuzzy processor with specialized fuzzy computations and 2) the dedicated fuzzy hardware for specific applications. Because the general-purpose fuzzy processor can be implemented quickly and applied flexibly, and dedicated fuzzy hardware requires long time for development, the general-purpose fuzzy processor-8051 microcontroller can be used. Nevertheless, there are also systemic uncertainties and disturbance in FLC controller. Because sliding-mode control (SMC) had been known as an effective approach to restrain the systemic uncertainties and disturbance, SMC algorithm was utilized. In order to achieve a robust control system, the microcontroller of the biochip system combining FLC and SMC algorithms optimally has been developed. Therefore, an OFSMC based on an 8051 microcontroller has been investigated numerically and experimentally in this charter. Hence, microfluidic manipulation in the microchannel of the biochip system based on OFSMC has been implemented by using an 8051 microcontroller.

The microfluidic manipulation based on the microcontroller has successfully been utilized to improve the reaction efficiency of molecular biology. First, it was used in DNA hybridization. There are two methods to improve the efficiency of the nucleic acid hybridization in this charter. The first method is to increase the velocity of the target nucleic acid molecules, which increases the effective collision into the probe molecules as the target molecules flow back and forth. The second method is to introduce the strain rates of the target mixture flow on the hybridization surface. This hybridization chip was able to increase hybridization signal 6-fold, reduce non-specific target-probe binding and background noises within 30 minutes, as compared to conventional hybridization methods, which may take from 4 hours to overnight. Second, it was used in DNA extraction. When serum existed in the fluid, the extraction efficiency of immobilized beads with solution flowing back and forth was 88-fold higher than that of free-beads. Third, it could be integrated in lab-on-a-chip. For the Tee-connected channels, it demonstrated the ability of manipulating the liquid drop from a first channel to a second channel, while simultaneously preventing flow into the third channel. Because there is a continuous airflow at the "outlet" during fluid manipulation, it can avoid contamination of the air source similar to the "laminar flow hook" in biological experiments.

The charter is organized as follows. In Section 2, we introduce the structure of the biochip control system. In Section 3, the fundamental knowledge of OFSMC and the model of the biochip system are introduced, and we address the OFSMC scheme and the associated

simulations. In Section 4, the OFSMC IC based on 8051 microprocessor is designed, and the results of the real-time experiment are presented. In Section 5, the efficiency improvement for the molecular biology reaction and DNA extraction by using OFSMC method are presented. Finally, the conclusion is given in Section 6.

2. Structure of the biochip control system

The structure of the biochip control system (Fig. 1) contains six parts: an air compressor, two flow controllers and two flowmeters, a flow-control chip, a biochip, photodiodes system, and a control-chip circuit system. One had designed a pneumatic device with planar structures for microfluidic manipulation (Chung, Jen, Lin, Wu & Wu, 2003). Pneumatic devices without any microfabricated electrodes or heaters, which will have a minimal effect on the biochemical properties of the microfluid by not generating electrical current or heat, are most suitable for μ TAS. A pneumatic structure possessing the ability of bi-directional pumping should be utilized in order to implement a pneumatic device which can control the movement of microfluid without valves or moving parts.

Fig. 1. Structure of the biochip control system.

The schematic diagram of the single pneumatic structure, which provides suction and exclusion by two inlets, is depicted in Fig. 2. When the air flows through inlet A only, it causes a low-pressure zone behind the triangular block and suction occurs in the vertical microchannel. Furthermore, when the air flows through inlet B only, the airflow is induced into the vertical microchannel to generate exclusion. The numerical and experimental results of the pressure and the stream tracers for the condition of the flow-control chip have been demonstrated (Marquardt, 1963). According to the principle of the flow-control chip, the microfluidic manipulation on the biochip is presented in this study by using OFSMC rules

with two flow controllers and two flowmeters. Since the biochip in the biochip system is a consumer, the photodiodes system should be utilized for sensing the feedback signals of the position of the reagent in the microchannel of the biochip. Hence, DNA extraction can be achieved in this study.

Fig. 2. Single pneumatic structure.

3. Design of the biochip control system

3.1 Design of optimal fuzzy sliding mode control

The biochip system of this design is shown in Fig. 1. If the biochip is DNA extraction chip, the extraction beads are immobilized on the channels. When the bio-fluidics does not flow the place without beads, the time of not extracting DNA can be reduced, and the extraction efficiency will also be improved. So the control of bio-fluidics' location is critical to DNA extraction (or hybridization) efficiency.

The biochip system depicted in Fig. 1 is a nonlinear system. Since the mathematical model of the flow-control mechanism and the microchannels in the biochip is a complicated nonlinear model, FLC design of the controller was utilized. The basic idea behind FLC is to incorporate the expert experience of a human operator in the design of the controller in controlling a process whose input-output relationship is described by a collection of fuzzy control rules (Altrock, Krause & Zimmermann, 1992). The heart of the fuzzy control rules is a knowledge base consisting of the so-called fuzzy IF-THEN rules involving linguistic variables rather than a complicated dynamic model. The typical architecture of a FLC, shown in Fig. 3, is comprised of four principal components: a fuzzification interface, a knowledge base, an inference engine, and defuzzification interface. The fuzzification interface has the effect of transforming crisp measured data into suitable linguistic values; it was designed first so that further fuzzy inferences could be performed according to the fuzzy rules (Polkinghorne, Roberts, Burns & Winwood, 1994). The heart of the fuzzification interface is the design of membership function. There are many kinds of membership

functions - Gaussian, trapezoid, triangular and so on - of the fuzzy set. In this paper, a triangular membership function was utilized, as shown in Figs. 4-5.

Fig. 3. Architecture of a fuzzy logic controller.

Fig. 4. Membership function-input variable (d) of photodiode detector.

Fig. 5. Membership function-output variable (z) of photodiode detector.
The overall fuzzy rules for the biochip system are defined as the following:

- IF d is NB then z_j is MPB
- IF d is NM then z_j is MPM
- IF d is NS then z_j is MPS
- IF d is ZE then z_j is M
- IF d is PS then z_j is MNS
- IF d is PM then z_j is MNM
- IF d is PB then z_j is MNB

where d is input variable of the photodiode signal, and z is output variable of the photodiode signal.

The inference engine is based on the compositional rule of inference with knowledge base for approximate reasoning suggested by Zadeh (Zadeh, 1965; Zadeh, 1968). An inference engine is the kernel of the FLC in modeling human decision making within the conceptual framework of fuzzy logic and reasoning. Hence, the fuzzification interface and fuzzy rules are designed completely before fuzzy reasoning. In this paper, since there are many structures of inference engine, fuzzy reasoning-Mamdani’s minimum fuzzy implication rule (MMFIR) method (Mamdani, 1977; Lee, 1990; Altrock, Krause & Zimmermann, 1992; Lin and Lee, 1999) was utilized. For simplicity, assume two fuzzy rules as follows:

R1: IF x is A1 and y is B1, then z is C1,

R2: IF x is A2 and y is B2, then z is C2.

Then the firing strengths α_1 and α_2 of the first and second rules may be expressed as

$$\alpha_1 = \mu_{A_1}(x_0) \wedge \mu_{B_1}(y_0) \text{ and } \alpha_2 = \mu_{A_2}(x_0) \wedge \mu_{B_2}(y_0),$$

where $\mu_{A_1}(x_0)$ and $\mu_{B_1}(y_0)$ indicate the degrees of partial match between the user-supplied data and the data in the fuzzy rule base.

In MMFIR fuzzy reasoning, the i th fuzzy control rule leads to the control decision

$$\mu_{C_i}(w) = \alpha_i \wedge \mu_{C_i}(w).$$

The final inferred consequent C is given by

$$\mu_C(w) = \mu_{C_1} \vee \mu_{C_2} = [\alpha_1 \wedge \mu_{C_1}(w)] \vee [\alpha_2 \wedge \mu_{C_2}(w)].$$

The fuzzy reasoning process is illustrated in Fig. 6.

Fig. 6. Fuzzy reasoning of MMFIR method.

Defuzzification is a mapping from a space of fuzzy control actions defined over an output universe of discourse into a space of crisp control actions. This process is necessary because fuzzy control actions cannot be utilized in controlling the plant for practical applications. Hence, the widely used center of area (COA) method, which generates the center of gravity of the possibility distribution of a control action, was utilized. In the case of a discrete universe, this method yields

$$z_{COA} = \frac{\sum_{j=1}^n \mu_C(z_j) z_j}{\sum_{j=1}^n \mu_C(z_j)} \tag{1}$$

where n is the number of quantization levels of the output, z_j is the amount of control output at the quantization level j , and $\mu_C(z_j)$ represents its membership degree in the output fuzzy set C .

The biochip system depicted in Fig. 1 is a nonlinear system that has been used as an application to study real world nonlinear control problems by different control techniques (Cheng & Li, 1998; Li & Shieh, 2000). The model of the biochip system is identified by ARX model, as

$$\begin{cases} X(k+1) = A_z X(k) + B_z u(k) \\ y(k) = C_z X(k) \end{cases} \quad (2)$$

where $X(k) \in R^n$ is the state variables of system, $u(k) \in R^m$ is the input voltage of the flow controller and $y(k) \in R^r$ is the assumed model output related to the position of the reagent in the microchannel of the biochip. The system is controllable and observable.

Sliding mode control's robust and disturbance-insensitive characteristics enable the SMC controller to perform well in systems with model uncertainty, disturbances and noises. In this paper, in addition to FLC controller, SMC controller was utilized to design the control input voltage of the flow controller. To design SMC controllers, a sliding function was designed first, and then enforced a system trajectory to enter sliding surface in a finite time. As soon as the system trajectory entered the sliding surface, they moved the sliding surface to a control goal. To sum up, there are two procedures of sliding mode, as shown in Fig. 7.

Fig. 7. Generation of sliding mode.

The proposed SMC controller was based on pole placement (Chang, 1999), since the sliding function could be designed by pole placement. Some conditions were set for the sliding vector design in the proposed sliding mode control:

1. $\text{Re}\{\lambda_i\} < 0$, $\alpha_j \in R$, $\alpha_j < 0$, $\alpha_j \neq \lambda_i$.
2. Any eigenvalue in $\{\alpha_1, \dots, \alpha_m\}$ is not in the spectrum of A_z .
3. The number of any repeated eigenvalues in $\{\lambda_1, \dots, \lambda_{n-m}, \alpha_1, \dots, \alpha_m\}$ is not greater than m , the rank of B_z .

where $\{\lambda_1, \lambda_2, \dots, \lambda_{n-m}\}$ are sliding-mode eigenvalues and $\{\alpha_1, \alpha_2, \dots, \alpha_m\}$ are virtual eigenvalues.

As proved by Sinswat and Fallside (Sinswat & Fallside, 1977), if the condition (3) in the above is established, the control system matrix $A_z - B_z K$ can be diagonalized as

$$A_z - B_z K = \begin{bmatrix} V \\ F \end{bmatrix}^{-1} \begin{bmatrix} \Phi_V & 0 \\ 0 & \Gamma_F \end{bmatrix} \begin{bmatrix} V \\ F \end{bmatrix} \quad (3)$$

where $\Phi_V = \text{diag}[\lambda_1, \lambda_2, \dots, \lambda_{n-m}]$, $\Gamma_F = \text{diag}[\alpha_1, \alpha_2, \dots, \alpha_m]$, and V and F are left eigenvectors with respect to Φ_V and Γ_F , respectively. Hence, Eq. (3) can be rewritten as

$$\begin{cases} V(A_z - B_z K) = \Phi_V V \\ F(A_z - B_z K) = \Gamma_F F \end{cases} \quad (4)$$

Rearrangement of Eq. (4) yields

$$FA_z - \Gamma_F F = (FB_z)K \quad (5)$$

According to Chang (Chang, 1999),

$$\text{rank}(FA_z - \Gamma_F F) = \text{rank}(F) \quad (6)$$

Since F contains m independent left eigenvectors, one has $\text{rank}(F) = m$. From Eqs. (5) and (6), it is also true that $\text{rank}(FA_z - \Gamma_F F) = \text{rank}((FB_z)K) = \text{rank}(F) = m$. In other words, FB_z is invertible. With the designed left eigenvector F above, the sliding function $S(k)$ is designed as

$$S(k) = FX(k) \quad (7)$$

The second step is the discrete-time switching control design. A different and much more expedient approach than that of Gao et al. (Gao, Wang & Homaifa, 1995) is adopted here. This approach is called the reaching law approach that has been proposed for continuous variable structure control (VSC) systems (Gao, 1990; Hung, Gao & Hung, 1993; Gao & Hung, 1993). This control law is synthesized from the reaching law in conjunction with a plant model and the known bounds of perturbations. For a discrete-time system, the reaching law is (Gao, Wang & Homaifa, 1995)

$$S(k+1) - S(k) = -qTS(k) - \varepsilon T \text{sgn}(S(k)) \quad (8)$$

where $T > 0$ is the sampling period, $q > 0$, $\varepsilon > 0$ and $1 - qT > 0$. Therefore, the switching control law for the discrete-time system is derived based on this reaching law. From Eq. (7) and pole-placement method, $S(k)$ and $S(k+1)$ can be obtained in terms of sliding vector F as,

$$\begin{cases} S(k) = FX(k) \\ S(k+1) = FX(k+1) = F(A_z - B_z K)X(k) + FB_z u(k) \end{cases} \quad (9)$$

where $K \in R^n$ is a gain matrix obtained by assigning n desired eigenvalues $\{\lambda_1, \dots, \lambda_{n-m}, \alpha_1, \dots, \alpha_m\}$ of $A - BK$. It follows that

$$S(k+1) - S(k) = F(A_z - B_z K)X(k) + FB_z u(k) - FX(k) \quad (10)$$

From Eqs. (8) and (10),

$$S(k+1) - S(k) = -qTS(k) - \varepsilon T \operatorname{sgn}(S(k)) = F(A_z - B_z K)X(k) + FB_z u(k) - FX(k)$$

Solving for $u(k)$ obtains the switching control law

$$u(k) = -(FB_z)^{-1} [F(A_z - B_z K)X(k) + (qT - 1)FX(k) + \varepsilon T \operatorname{sgn}(FX(k))] \quad (11)$$

In order to achieve the output tracking control, a reference command input $r(k)$ is introduced into the system by modifying the state feedback control law $u_p(k) = -KX(k)$ with pole-placement design method (Franklin, Powell & Workman, 1998) to become

$$u_p(k) = N_u r(k) - K(X(k) - N_x r(k)) \quad (12)$$

where

$$\begin{bmatrix} N_u \\ N_x \end{bmatrix} = \begin{bmatrix} A_z - I & B_z \\ C_z & 0 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ I \end{bmatrix} \quad (13)$$

The proposed SMC input, based on Eq. (13), is assumed to be

$$u_s(k) = u_p(k) + u = N_u r(k) - K(X(k) - N_x r(k)) + u \quad (14)$$

Substituting Eq. (11) into (14) gives the proposed SMC input as

$$u_s(k) = N_u r(k) - K(X(k) - N_x r(k)) - (FB_z)^{-1} [F(A_z - B_z K)X(k) + (qT - 1)FX(k) + \varepsilon T \operatorname{sgn}(FX(k))] \quad (15)$$

The pole-placement SMC design method utilizes the feedback of all the state variables to form the desired sliding vector. In practice, not all the state variables are available for direct measurement. Hence, it is necessary to estimate the state variables that are not directly measurable.

In practice, a discrete linear time-invariant system sometimes has system disturbances and measurement noise. Hence, linear quadratic estimator (LQE) will be applied here to estimate optimal states in having system disturbances and measurement noise.

According to Eq. (2), consider a system model as

$$\begin{cases} X(k+1) = A_z X(k) + B_z u(k) + Gv(k) \\ y(k) = C_z X(k) + \omega(k) \end{cases} \quad (16)$$

where $X(k) \in R^n$ is the state variable, $u(k) \in R^m$ is the control input voltage, $y'(k) \in R^r$ is the assumed plant output related to the XY stage position, and $v(k) \in R^n$ and $\omega(k) \in R^r$ are system disturbances and measurement noise with covariances $E[\omega\omega^T] = Q$, $E[vv^T] = R$ and $E[\omega v^T] = 0$.

The objective of LQE is to find a vector $\hat{X}(k)$ which is an optimal estimation of the present state $X(k)$. Here "optimal" means the cost function

$$J = \lim_{T \rightarrow \infty} E \left\{ \int_0^T (X^T Q X + u^T R u) dt \right\} \tag{17}$$

is minimized. The solution is the estimator as

$$\begin{cases} \hat{X}(k+1) = A_z \hat{X}(k) + B_z u(k) + K_f (y(k) - C_z \hat{X}(k)) \\ \hat{y}(k) = C_z \hat{X}(k) \end{cases} \tag{18}$$

where K_f is the “optimal Kalman” gain $K_f = PC_z^T R^{-1}$ and P is the solution of the algebraic Riccati equation

$$A_z P + P A_z^T - P C_z^T R^{-1} C_z P + Q = 0 \tag{19}$$

FLC, SMC, and LQE were combined into the so called optimal fuzzy sliding-mode control (OFSMC) and utilized to control input voltage of the flow controller. The OFSMC block diagram with LQE is shown in Fig. 8.

Fig. 8. OFSMC block diagram.

In the biochip system, the photodiode system provided the position feedback signal for FLC and LQE. Then, the FLC could use the position feedback signals to generate the input signals for SMC. And the LQE could estimate optimal states in having system disturbances and measurement noise for SMC by the position feedback signals. Hence, the SMC with FLC and LQE could implement the microfluidic manipulation very well and robustly. The performance of the OFSMC would be explained in detail by simulation and experimental results, which are presented in Section 3.2.

3.2 Simulation of OFSMC

This section deals with a system model described by Eq. (2) and defines a reference command input $r(k)$, which is an input voltage of the flow controller by fuzzy controller with designed photodiode signals. The pole-placement algorithm described in Section 3.1 is utilized to determine a sliding vector. In this study, Ackermann's Formula is used to determine the pole-placement feedback gain matrix K . In practice, the fact that not all state variables are available for direct measurement results in the necessity to estimate the state variables that are not directly measurable. Hence, the full-order state observer designed by Ackermann's Formula and LQE will be utilized in this study.

In order to achieve the biochip application, the microfluidic reagent has to be manipulated to flow back and forth in the central zone of the microchannel between the PD1 and PD2, shown in Fig. 1. During the simulations, the external disturbance would be added in system plant. Figs. 9 and 10 show the simulations of the biochip system model at 2 Hz of back and forth flowing based on FLC, and fuzzy sliding mode control (FSMC), respectively, with the full-order estimator (FOE), and OFSMC by using the MATLAB and Simulink. In Figure 9, the blue solid lines represent reference command input whereas the red dotted lines, the green dash-dot lines and the magenta dashed line are the system output based on FLC, FSMC with FOE and OFSMC respectively. Every turn of the curve represents a reversal of the flowing reagent during its back and forth flow in the microchannel on the biochip. Fig. 10 is the error performance of the simulation results of biochip system model based on the three controllers.

Fig. 9. Simulation results of biochip system model based on FLC, FSMC with the FOE and OFSMC at 2 Hz. The blue solid lines represent reference command input whereas the red dotted lines, the green dash-dot lines and the magenta dashed line are the system output based on FLC, FSMC with FOE and OFSMC respectively.

Increasing emphasis on the mathematical formulation and measurement of control system performance can be found in recent literature on modern control. Therefore, as an always-

positive number or zero, the performance index that can be calculated or measured and used to evaluate the system's performance is usually utilized. The best system is defined as the system that minimizes this index. In this study, integrated absolute error (IAE) that is often of practical significance is used as the performance index and is expressed as

$$IAE = \int_0^T |e(t)| dt \quad (20)$$

where $e(t)$ is a error function of the plant and T is a finite time. In addition to IAE, integral of time multiplied by absolute error (ITAE) that provides the performance index of the best sensitivity is expressed as

$$ITAE = \int_0^T t |e(t)| dt \quad (21)$$

where $e(t)$ is a error function of the plant and T is a finite time. Using the above two methods, the performance of the system will be evaluated exactly.

In molecular biology applications, increasing the velocity of the target nucleic acid molecules increases the number of effective collision into the probe molecules as the target molecules flow back and forth, which will ultimately increase the efficiency of biochemical reaction obviously. Therefore, according to the issue, the performance of the simulation results with the three control rules as the target molecules flowing back and forth at 0.2, 0.5, 1 and 2 Hz would be presented in Table 1.

Fig. 10. Error performance of simulation results of biochip system model based on FLC, FSMC with the FOE and OFSMC at 2 Hz. The red dotted lines, the green dash-dot lines and the magenta dashed line are the system output based on FLC, FSMC with FOE and OFSMC respectively.

The overshoot means the reagent is out of the central zone as it is manipulated. Here, out of the length of the central zone is defined as overshoot value. And the error performance of the simulation results were also evaluated by using IAE and ITAE indices and the results are shown in Tables 2 and 3. The following conclusions can be arrived at from the analysis of the simulation results from Figs. 9 and 10, and Tables 1 to 3.

<div>Frequency (Hz)</div> <div>Overshot (mm)</div>	FLC	FSMC	OFSMC
0.2	8.0	1.1	0.5
0.5	10.0	1.3	0.7
1	11.5	1.7	0.9
2	12.0	1.8	1.1

Table 1. Performances of the simulation results with FLC, FSMC, and OFSMC control rules

<div>Frequency (Hz)</div> <div>IAE</div>	FLC	FSMC	OFSMC
0.2	3061	1106	73
0.5	2888	1084	100
1	2889	1279	162
2	2892	1824	301

Table 2. IAE index of the control systems with FLC, FSMC, and OFSMC control rules

<div>Frequency (Hz)</div> <div>ITAE</div>	FLC	FSMC	OFSMC
0.2	5438	2063	118
0.5	5514	2023	170
1	5712	2473	303
2	5814	3529	570

Table 3. ITAE index of the control systems with FLC, FSMC, and OFSMC control rules

1. According to Figs. 9, 10, and the values of IAE and ITAE (Table 1), the biochip system model based on OFSMC controller at 2 Hz performs better than that based on FLC, or FSMC controller with FOE. In addition, the performances of the biochip system model based on OFSMC controller at 0.2, 0.5, and 1 Hz are obviously better than the other two, according to Tables 1 to 3. Therefore, the OFSMC controller can perform well in the biochip system with disturbances.
2. It is certain that the OFSMC control method is capable of manipulating the position of the reagent in the microchannel on the biochip robustly and successfully. The experimental results of microfluidic manipulation on biochip system with OFSMC controller based on 8051 microprocessor are shown in Section 4.

4. Experimental results of OFSMC

The control block diagram of the biochip system with OFSMC controller described in Section 2 and 3 is shown in Fig. 8. In order to provide a quick and useful product for non PC-based systems, the microfluidic manipulation is implemented by 8051 microprocessor in this study. And the A/D and D/A chips were utilized to convert the photodiode or flowmeter feedback analog signals into digital signals for the microprocessor as well as to convert digital signals into analog signals for the flow controller. Then, the circuit of the photodiode-signal process should be designed. Assembly language was utilized to program the OFSMC control rules to embed into 8051 microprocessor with flow chart of the program shown in Fig. 11. The experimental results of microfluidic manipulation on biochip system with OFSMC controller based on 8051 microprocessor are shown in Fig. 12, where the volume of reagent used is 94 μL . The reagent on the biochip system was controlled excellently to flow back and forth at 2 Hz, because the overshoot of the control performance was very small and the control system was very stable. The experimental results of the control performance with FLC, FSMC, and OFSMC control rules are shown in Fig. 13.

According to Fig. 13, the microfluidic manipulation with FLC control rule can only be implemented to flow back and forth at 0.2 Hz, and the overshoot of the performance is -10, which means the reagent could not be manipulated between the PD1 and PD2. Either it was pushed out of the biochip, or it was manipulated under 1 cm length of the undershoot at 0.2 Hz of back and forth flowing. In addition, according to the results of the performance with FSMC control rule, the overshoot became larger and larger by increasing the frequency of back and forth flowing. Compared to FLC and FSMC control rule, the overshoots of the performance with OFSMC control rule were the least of the three control rules and the performance was the most stable and the best of the three at all frequencies of back and forth flowing. The microfluidic manipulation on biochip system with OFSMC rules can keep flowing back and forth at 2 Hz within 1 h while the other two can not.

Since the experimental and simulation results are in good agreement, it could be concluded that the control performance with OFSMC control rule was better than that with FLC and FSMC. Compared to FLC and FSMC, it was more successful to overcome the variable parameters and nonlinear model to achieve a better microfluid management with OFSMC control rule when using different biochip for every time. Therefore, it is certain that the OFSMC control method is capable of manipulating the position of the reagent in the microchannel on the biochip robustly and successfully.

Fig. 11. Flow chart of the program.

Fig. 12. Experimental results of microfluidic manipulation at 2 Hz (A period from (A) to (D)).

Fig. 13. Experimental results of microfluidic manipulation with FLC, FSMC, and OFSMC control rules.

5. Biomedical application results

According to the experimental results given in Section 4, the microfluidic manipulation based on the microcontroller could be utilized in biotechnology, as it successfully improved the efficiency of the biochemical reaction. First, it was used in DNA hybridization. There are

two methods to improve the efficiency of the nucleic acid hybridization in this chapter. The first method is to increase the velocity of the target nucleic acid molecules, which increases the effective collision into the probe molecules as the target molecules flow back and forth. The second method is to introduce the strain rates of the target mixture flow on the hybridization surface. This hybridization chip was able to increase hybridization signal 6-fold, reduce non-specific target-probe binding and background noises within 30 minutes, as compared to conventional hybridization methods, which may take from 4 hours to overnight. Second, it was used in DNA extraction. In this section, DNA extraction from Chung, Wen and Lin (Chung, Wen & Lin, 2007) is introduced.

The microfluidic DNA extraction chip was designed and fabricated onto a polymethylmethacrylate (PMMA) substrate with flow channels. Machined and immobilized-beads PMMA substrate and blank PMMA were bonded together to form the device. The beads used for DNA extraction was obtained from Magic Bead Inc. (USA). A plasma generator was used to perform the surface treatment. Plasma source gas consisting of a mixture of ammonia and oxygen was used to activate the surface of PMMA substrate.

Escherichia coli (*E. coli*) was cultured in 5 ml of LB medium (NaCl: 10 g/l, Tryptone: 10 g/l, yeast extract: 5 g/l) in 15 ml tubes at 37°C and 225 rpm. After 16 h, the optical density (OD) of the culture was measured in a spectrophotometer (U-2100, Hitachi, Japan). The number of *E. coli* cells or the amount of DNA was calculated from an OD versus cell number. The culture was then diluted by distilled water to obtain varying numbers (10² -10⁵) of *E. coli* cells per micro liter. DNA was extracted from the blood of one of the members of the group using the microchip. Whole blood was directly used without any pretreatment.

The sample flowed forward and backward with the immobilized beads at a frequency of 1 Hz inside the channel. *E. coli* cells were treated with a buffer (B1+B2, Magic Bead, USA) to lyse the cells and to release the DNA. The DNA extracted using the microchips was amplified by PCR (Polymerase Chain Reaction). The forward and reverse primers were 5'-CAGGATTAGATACCCITGGTAG-3' and 5'-TTCCCCTACGGTTACCTTGTT-3', respectively. The PCR condition was: one cycle of 5 min at 95°C, 40 cycles of 30 s at 95°C, 40 s at 58°C and 40 s at 72°C, and one cycle of 10 min at 72°C. The PCR products were analyzed qualitatively in a Mupid-2 electrophoresis equipment (Advance, Japan) and quantitatively in an Agilent 2100 bioanalyzer (Agilent Technologies, USA).

The extraction efficiencies of *E. coli* cell number in the whole blood were tested and are shown in Fig. 18a. It showed that the free beads could efficiently extract DNA as the number of *E. coli* cells was higher than 5×10⁴, but could hardly extract DNA as the number was smaller than 10⁴ in 25 µl of whole blood. And for the immobilized beads, the corresponding boundary number of *E. coli* cells for efficient and hard extractions of DNA were 2×10² and 10², respectively. After the analysis in the bioanalyzer, the results were as shown in Fig. 18b. When the number of *E. coli* cells was 2×10² to 10⁴ in 25 µl of whole blood, the extraction efficiency of immobilized beads with solution flowing back and forth was about 600-fold larger than that of free beads.

6. Conclusions

In biometric and biomedical applications, an important issue for miniaturization and integration is microfluid management. This chapter introduced the optimal fuzzy sliding-mode control (OFSMC) design based on the 8051 microprocessor and the complete microfluidic manipulated system implementation of biochip system with a pneumatic pumping actuator, two feedback-signal photodiodes and flowmeters for better microfluid

management. The newly developed microfluid management technique was successfully utilized to improve the reaction and extraction efficiency of a biochemical reaction.

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8. References

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The main objective of this monograph is to present a broad range of well worked out, recent application studies as well as theoretical contributions in the field of sliding mode control system analysis and design. The contributions presented here include new theoretical developments as well as successful applications of variable structure controllers primarily in the field of power electronics, electric drives and motion steering systems. They enrich the current state of the art, and motivate and encourage new ideas and solutions in the sliding mode control area.

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