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Dynamics and Control for Nonholonomic Mobile Modular Manipulators

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1. Introduction

The development of a robot requires that it be able to adopt as many configurations as possible using limited modules, so as to allow the construction of new types of robots without redesign and remanufacturing. Traditionally, modular manipulators are mounted on a fixed base whose mobility is constrained. However, with the development of industry and technology, such modular manipulators as mounted on fixed bases can not meet some practical requirements any more. An intelligent and autonomous mobile manipulator, which can fulfil some operations without human interference, has become an active research topic recently since it has many potential applications such as in modern factories for transporting materials, in dangerous fields for dismantling bombs or moving nuclear infected objects, in modern families for doing housework, as well as in the public places for city maintenance.

In this chapter, a nonholonomic mobile platform is attached to the modular manipulator in order to increase workspace of the entire robot. Building up the dynamic model for a nonholonomic mobile modular manipulator is a challenging task due to the interactive motions between the modular manipulator and the mobile platform, as well as the nonholonomic constraints of the mobile platform. Also a trajectory following task becomes even more complex and difficult to achieve.

Such conventional control strategies as computed-torque control require precise apriori knowledge of the dynamic parameters for the controlled system. However, in practical applications, it is almost impossible to obtain exact dynamic parameters for a mobile modular manipulator because of such uncertainties as complex nonlinear frictions, flexibilities of the joints and links, payload variations, and terrain irregularities. Robust control techniques provide a natural rejection to external disturbances, which are provided by a high-frequency commuted control action that constrains the error trajectories to stay on the sliding surface. Classical sliding mode control law adopts sign functions and the caused chattering may do harm to the robots. Adaptive control technique does not rely on precise apriori knowledge of dynamic parameters and it can suppress such errors as caused by parameter uncertainties by online adjusting dynamic parameters. Furthermore, adaptive control can counteract the negative influence of high-frequency switching caused by robust control because its action has naturally smooth time behaviour.

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In related work on modular robots, the modular robot concept could be traced back to the 1970's (Will & Grossman, 1975). In early modular robot research, the emphasis was put on the structure design of self-organizing, self-reconfigurable, self-assembling, and self-repairable modular robots (Fukuda et al., 1989; Tomita et al., 1999). Kinematic and dynamic analysis as well as trajectory planning became another active topic in the past decades (Chen & Yang, 1998; Fei et al., 2001). In recent years, the scholars had turned their attentions to trajectory following control for modular manipulators (Melek & Goldenberg, 2003; Shen et al., 2002; Stoy et al., 2002). Parameter identification and vibration control for a 9-DOF reconfigurable modular manipulator were investigated by authors in (Li et al., 2004a).

Regarding to literatures on mobile manipulators, mobile manipulators were exploited to install and remove aircraft warning spheres (Campos et al., 2002), to polish aircraft canopy (Jamisola et al., 2002), to organize furniture in a room (Rus et al., 1995), and to collectively transport a single palletized load (Stilwell & Bay, 1993). A great deal of research activities can be found on motion planning of mobile manipulators (Carriker et al., 1991; Chitta & Ostrowski, 2002; Nagatani et al., 2002). Several kinematic and dynamic modelling methods were presented for mobile manipulators in the past decade, such as the Kane's method (Tanner & Kyriakopoulos, 2001), the Newton-Euler method (Chung & Velinsky, 1999) and the Lagrange method (Li & Liu, 2004b; Liu & Li, 2005a; Yu & Chen, 2002). Tip-over analysis and prevention attracted numerous scholars and several tip-over stability criteria were defined, such as the potential energy stability level (Ghasempoor & Sepehri, 1995), the force-angle stability measure (Papadopoulos & Rey, 1996), the zero moment point criterion (Furuno et al., 2003), and the criterion based on supporting forces (Li & Liu, 2005b). Extensive literatures can be found on control of mobile manipulators. Dynamic characteristics between the mobile platform and the onboard manipulator were investigated (Yamamoto & Yun, 1996). A robust control method was developed to eliminate the harmful effect of the wheel slip on the tracking performance of a spatial mobile manipulator (Chung & Velinsky, 1999). A homogeneous kinematic stabilization strategy and an adaptive control scheme were combined for mobile manipulator control without any knowledge of the system dynamic model (Colbaugh, 1998). Neural network and fuzzy logic control for mobile manipulators were also studied by authors (Li & Liu, 2005c, 2005d, 2006a, 2006b).

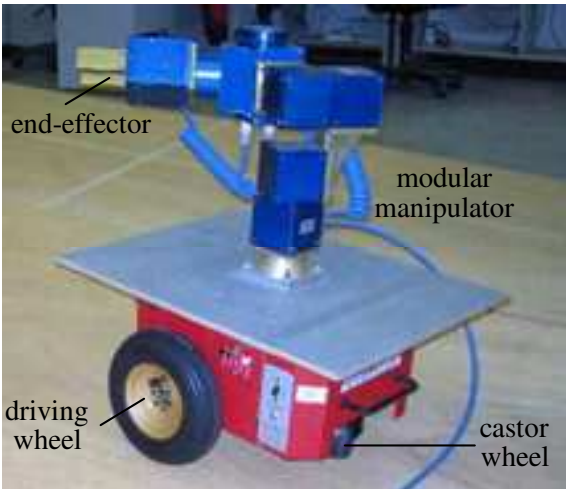
In the previous research work, modeling for the mobile platform and for the manipulator was usually carried out separately and control for nonholonomic mobile robots was mostly limited to kinematic velocity control, while few work on dynamic torque control. The interactive motions between the manipulator and the mobile platform made the models established inaccurate, which then affected the control result. Most of the present controllers were designed in joint space, but few in task space (Ge et al., 1997). However, in practical applications, the end-effector of a robot is usually specified to fulfil some operations.

This chapter is organized as follows: an integrated modelling method is proposed considering nonholonomic constraints and interactive motions in Section 2. In Section 3, a robust adaptive controller is designed in task space to control the end-effector to follow desired spatial trajectories. Simulations are conducted on a real mobile modular manipulator, and a comparison is made with the conventional model-based controller in Section 4. Section 5 gives some concluding remarks. Some suggested ideas for future research work are presented in the last section.

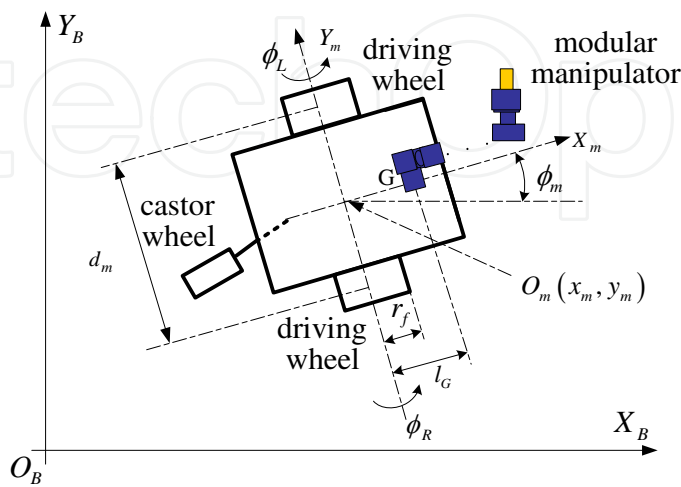
2. An Integrated Modelling Method

2.1 Kinematics analysis

A mobile modular manipulator is normally composed of an m -wheeled holonomic or nonholonomic mobile platform and an n -DOF onboard modular manipulator, as shown in Fig. 1(a). In this chapter, we analyze the 3-wheeled nonholonomic mobile platform, which has two driving wheels and one castor wheel. The two driving wheels are coaxial and mounted in front of the platform and the castor wheel in the rear is orientable with respect to the cart. The robot is assumed to move on a horizontal plane; then the motion of the mobile platform can be illustrated as shown in Fig. 1(b). An arbitrary inertial base frame $O_B X_B Y_B Z_B$ is fixed on the motion plane, while a frame $O_m X_m Y_m Z_m$ is attached to the mobile platform. In frame $O_m X_m Y_m Z_m$, $O_m(x_m, y_m)$ is selected as the midpoint of the line segment connecting the two driving-wheel centres; $O_m Y_m$ is along the coaxial of the two driving wheels; $O_m X_m$ is perpendicular to $O_m Y_m$ and passes through O_m . The heading angle ϕ_m determines the posture of the mobile platform.



(a) Prototype



(b) coordinate system definition

Fig. 1. Prototype and coordinate system definition for a mobile modular manipulator.

Since the modular manipulator is mounted on the mobile platform, it can be assumed that the relative position of frame $O_m X_m Y_m Z_m$ and the frame $O_0 X_0 Y_0 Z_0$ is steady; here $O_0 X_0 Y_0 Z_0$ is the frame attached to the 1st module of the modular manipulator. Therefore the mobile platform can be treated as a special module added to the bottom of the modular manipulator, which can both move on the motion plane and rotate about the vertical axis. From Fig. 1(b), transformation matrices between the frames $O_B X_B Y_B Z_B$, $O_m X_m Y_m Z_m$ and $O_0 X_0 Y_0 Z_0$ are given by

$${}^B_m T = \begin{bmatrix} \cos \phi_m & -\sin \phi_m & 0 & x_m \\ \sin \phi_m & \cos \phi_m & 0 & y_m \\ 0 & 0 & 1 & r_f \\ 0 & 0 & 0 & 1 \end{bmatrix}, \quad {}^m_0 T = \begin{bmatrix} 1 & 0 & 0 & l_G \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & h_G \\ 0 & 0 & 0 & 1 \end{bmatrix} \quad (1)$$

where r_f is the radius for the driving wheels; l_G and h_G denote offsets of the 1st module with respect to $O_m X_m Y_m Z_m$ along $O_m X_m$, $O_m Y_m$ respectively.

By defining the link four parameters according to the Denavit-Hartenberg (D.H.) notation (de Wit et al., 1996), the transformation matrix between two adjacent links for the modular manipulator can be derived as follows

$${}^{i-1}_i T = \begin{bmatrix} \cos q_i & -\sin q_i & 0 & l_i \\ \sin q_i \cos \alpha_i & \cos q_i \cos \alpha_i & -\sin \alpha_i & -d_i \sin \alpha_i \\ \sin q_i \sin \alpha_i & \cos q_i \sin \alpha_i & \cos \alpha_i & d_i \cos \alpha_i \\ 0 & 0 & 0 & 1 \end{bmatrix} \quad (2)$$

where l_i, α_i, d_i, q_i are D.H. parameters.

Therefore the transformation matrix of the end effector with respect to the inertial base frame can be derived by

$${}^B_n T = {}^B_0 T {}^0_1 T {}^1_2 T \cdots {}^{n-1}_n T \quad (3)$$

Hence, the position vector \bar{p}_e and the posture vectors $\bar{n}, \bar{o}, \bar{a}$ of the end-effector with respect to the inertial base frame can be derived, which can be observed from the fourth column and the first three columns of the matrix ${}^B_n T$ correspondingly.

However, it is inconvenient to describe the posture of the end-effector using these posture vectors with nine parameters. The posture can be determined by only three independent parameters using Z-Y-Z Euler angles ϕ, θ and ψ . Relationship between the posture vectors and the Euler angles can be described as follows

$$\begin{aligned} \phi &= a \tan 2(a_y, a_x), \\ \theta &= a \tan 2(o_z, -n_z), \\ \psi &= a \tan 2(a_x \cos \phi + a_y \sin \phi, a_z) \end{aligned} \quad (4)$$

2.2 Nonholonomic constraints and Jacobian

The motion of a mobile modular manipulator during a very short time interval $[t^i \ t^{i+1}]$ is shown in Fig. 2. l_r , d_r and d_m represent the distances from the fixed bar to $O_m Y_m$, from the fixed bar to the horizontal axis of the castor wheel, and that between the two driving wheels respectively; r_i and θ_i are steering radius and yaw angle during the time interval; ϕ_L , ϕ_R , ϕ_r and β_r denote rotating angles of the left driving wheel, the right driving wheel, the castor wheel around its horizontal axis, and that around the fixed bar respectively; ΔS_L , ΔS_R and ΔS_m represent advances of the left driving wheel, the right driving wheel, and the mobile platform during the time interval correspondingly.

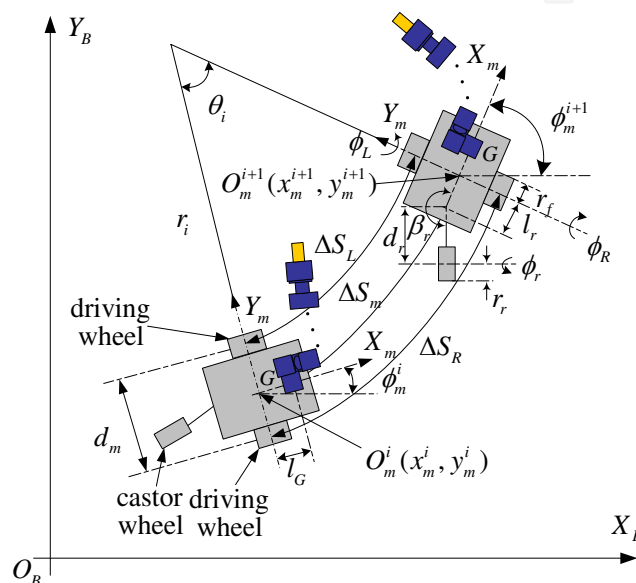


Fig. 2. Nonholonomic constraints derivation.

Let the time interval $\Delta t = t^{i+1} - t^i \rightarrow 0$, on the assumption of low speeds, velocities during this time interval can be treated as constants. Then,

$$\begin{cases}
 \Delta x_m^i = |O_m^i O_m^{i+1}| \cdot \cos \phi_m^i & (a) \\
 \Delta y_m^i = |O_m^i O_m^{i+1}| \cdot \sin \phi_m^i & (b) \\
 \Delta S_L = \theta_i \cdot \left(r_i - \frac{d_m}{2} \right) & (c) \\
 \Delta S_R = \theta_i \cdot \left(r_i + \frac{d_m}{2} \right) & (d) \\
 \Delta S_m \approx |O_m^i O_m^{i+1}| & (e) \\
 \theta_i \approx \Delta \phi_m^i & (f) \\
 r_i = \frac{\Delta S_m}{\theta_i} & (g)
 \end{cases} \quad (5)$$

where $\Delta x_m^i = x_m^{i+1} - x_m^i$, $\Delta y_m^i = y_m^{i+1} - y_m^i$ and $\Delta \phi_m^i = \phi_m^{i+1} - \phi_m^i$.

Substituting Eq. 5(f, g) into Eq. 5(c, d) yields

$$\begin{cases} \Delta S_L = \Delta S_m - \frac{d_m}{2} \cdot \Delta \phi_m^i & (a) \\ \Delta S_R = \Delta S_m + \frac{d_m}{2} \cdot \Delta \phi_m^i & (b) \end{cases} \quad (6)$$

Adding Eq. 6(a) and Eq. 6(b) yields

$$\Delta S_m = \frac{\Delta S_L + \Delta S_R}{2} \quad (7)$$

Subtracting Eq. 6(a) for Eq. 6(b) yields

$$\Delta \phi_m = - \frac{\Delta S_L - \Delta S_R}{d_m} \quad (8)$$

Substituting Eq. 7 and Eq. 5(e) into Eq. 5(a,b) yields

$$\begin{cases} \Delta x_m^i = \frac{\cos \phi_m^i}{2} \cdot (\Delta S_L + \Delta S_R) \\ \Delta y_m^i = \frac{\sin \phi_m^i}{2} \cdot (\Delta S_L + \Delta S_R) \end{cases} \quad (9)$$

From Eq. 8 and Eq. 9,

$$\begin{cases} \dot{x}_m^i = \lim_{\Delta t \rightarrow 0} \left(\frac{\Delta x_m^i}{\Delta t} \right) = \frac{\cos \phi_m^i}{2} \cdot \left[\lim_{\Delta t \rightarrow 0} \left(\frac{\Delta S_L}{\Delta t} \right) + \lim_{\Delta t \rightarrow 0} \left(\frac{\Delta S_R}{\Delta t} \right) \right] = \frac{\cos \phi_m^i}{2} \cdot (\dot{S}_L + \dot{S}_R) \\ \dot{y}_m^i = \lim_{\Delta t \rightarrow 0} \left(\frac{\Delta y_m^i}{\Delta t} \right) = \frac{\sin \phi_m^i}{2} \cdot \left[\lim_{\Delta t \rightarrow 0} \left(\frac{\Delta S_L}{\Delta t} \right) + \lim_{\Delta t \rightarrow 0} \left(\frac{\Delta S_R}{\Delta t} \right) \right] = \frac{\sin \phi_m^i}{2} \cdot (\dot{S}_L + \dot{S}_R) \\ \dot{\phi}_m^i = \lim_{\Delta t \rightarrow 0} \left(\frac{\Delta \phi_m^i}{\Delta t} \right) = -\frac{1}{d_m} \cdot \left[\lim_{\Delta t \rightarrow 0} \left(\frac{\Delta S_L}{\Delta t} \right) - \lim_{\Delta t \rightarrow 0} \left(\frac{\Delta S_R}{\Delta t} \right) \right] = -\frac{1}{d_m} \cdot (\dot{S}_L - \dot{S}_R) \end{cases} \quad (10)$$

Equation 10 always holds during the course of plane motion, so the superscript “i” can be omitted. Considering that $\dot{S}_L = r_f \cdot \dot{\phi}_L$ and $\dot{S}_R = r_f \cdot \dot{\phi}_R$, we can obtain

$$\begin{cases} \dot{x}_m = \frac{r_f \cdot \cos \phi_m}{2} \cdot \dot{\phi}_L + \frac{r_f \cdot \cos \phi_m}{2} \cdot \dot{\phi}_R & (a) \\ \dot{y}_m = \frac{r_f \cdot \sin \phi_m}{2} \cdot \dot{\phi}_L + \frac{r_f \cdot \sin \phi_m}{2} \cdot \dot{\phi}_R & (b) \\ \dot{\phi}_m = -\frac{r_f}{d_m} \cdot \dot{\phi}_L + \frac{r_f}{d_m} \cdot \dot{\phi}_R & (c) \end{cases} \quad (11)$$

In the same way, we can obtain

$$\begin{cases} \dot{\phi}_r = \frac{\cos(\beta_r + \phi_m)}{r_r} \cdot \dot{x}_m + \frac{\sin(\beta_r + \phi_m)}{r_r} \cdot \dot{y}_m - \frac{l_r \sin \beta_r}{r_r} \cdot \dot{\phi}_m \\ \dot{\beta}_r = \frac{\sin(\beta_r + \phi_m)}{d_r} \cdot \dot{x}_m - \frac{\cos(\beta_r + \phi_m)}{d_r} \cdot \dot{y}_m - \frac{d_r - l_r \cos \beta_r}{d_r} \cdot \dot{\phi}_m \end{cases} \quad (12)$$

Adding Eq. 11(a) multiplied by $\cos \phi_m$ to Eq. 11(b) multiplied by $\sin \phi_m$ yields

$$\cos \phi_m \cdot \dot{x}_m + \sin \phi_m \cdot \dot{y}_m = \frac{r_f}{2} \cdot \dot{\phi}_L + \frac{r_f}{2} \cdot \dot{\phi}_R \quad (13)$$

Adding Eq. 11(c) multiplied by $\frac{d_m}{2}$ to Eq. 13 yields

$$\cos \phi_m \cdot \dot{x}_m + \sin \phi_m \cdot \dot{y}_m + \frac{d_m}{2} \cdot \dot{\phi}_m - r_f \cdot \dot{\phi}_R = 0 \quad (14)$$

Subtracting Eq. 11(c) multiplied by $\frac{d_m}{2}$ from Eq. 13 yields

$$\cos \phi_m \cdot \dot{x}_m + \sin \phi_m \cdot \dot{y}_m - \frac{d_m}{2} \cdot \dot{\phi}_m - r_f \cdot \dot{\phi}_L = 0 \quad (15)$$

Subtracting Eq. 11(b) multiplied by $\cos \phi_m$ from Eq. 11(a) multiplied by $\sin \phi_m$ yields

$$\sin \phi_m \cdot \dot{x}_m - \cos \phi_m \cdot \dot{y}_m = 0 \quad (16)$$

From Eq. 12 and Eq. 14-16, we have

$$\begin{bmatrix} \cos \phi_m & \sin \phi_m & -\frac{d_m}{2} & 0 & 0 & -r_f & 0 \\ \cos \phi_m & \sin \phi_m & \frac{d_m}{2} & 0 & 0 & 0 & -r_f \\ \sin \phi_m & -\cos \phi_m & 0 & 0 & 0 & 0 & 0 \\ \cos(\beta_r + \phi_m) & \sin(\beta_r + \phi_m) & -l_r \sin \beta_r & -r_r & 0 & 0 & 0 \\ \sin(\beta_r + \phi_m) & -\cos(\beta_r + \phi_m) & l_r \cos \beta_r - d_r & 0 & -d_r & 0 & 0 \end{bmatrix} \cdot \begin{bmatrix} \dot{x}_m \\ \dot{y}_m \\ \dot{\phi}_m \\ \dot{\phi}_r \\ \dot{\beta}_r \\ \dot{\phi}_L \\ \dot{\phi}_R \end{bmatrix} = 0 \quad (17)$$

in short $A(\xi) \cdot \dot{\xi} = 0$; here $\xi = [x_m \ y_m \ \phi_m \ \phi_r \ \beta_r \ \phi_L \ \phi_R]^T$.

From Eq. 11 and Eq. 12, we can easily obtain

$$\begin{bmatrix} \dot{x}_m \\ \dot{y}_m \\ \dot{\phi}_m \\ \dot{\phi}_r \\ \dot{\beta}_r \\ \dot{\phi}_L \\ \dot{\phi}_R \end{bmatrix} = \begin{bmatrix} \frac{r_f \cdot \cos \phi_m}{2} & \frac{r_f \cdot \cos \phi_m}{2} \\ \frac{r_f \cdot \sin \phi_m}{2} & \frac{r_f \cdot \sin \phi_m}{2} \\ -\frac{r_f}{d_m} & \frac{r_f}{d_m} \\ \frac{r_f (d_m \cos \beta_r + 2l_r \sin \beta_r)}{2d_m r_r} & \frac{r_f (d_m \cos \beta_r - 2l_r \sin \beta_r)}{2d_m r_r} \\ \frac{r_f (d_m \sin \beta_r + 2d_r - 2l_r \cos \beta_r)}{2d_m d_r} & \frac{r_f (d_m \sin \beta_r - 2d_r + 2l_r \cos \beta_r)}{2d_m d_r} \\ 1 & 0 \\ 0 & 1 \end{bmatrix} \cdot \begin{bmatrix} \dot{\phi}_L \\ \dot{\phi}_R \end{bmatrix} \quad (18)$$

in short $\dot{\xi} = S(\xi) \cdot [\dot{\phi}_L \ \dot{\phi}_R]^T$.

Substituting Eq. 18 into Eq. 17 yields

$$A(\xi) \cdot S(\xi) = 0 \quad (19)$$

Let $x = [p_x \ p_y \ p_z \ \phi \ \theta \ \psi]^T$ and $q = [\phi_L \ \phi_R \ q_1 \ q_2 \ \dots \ q_n]^T$ be coordinates in task space and joint space respectively; and define $\varsigma = [\xi^T \ q_1 \ \dots \ q_n]^T$ as extended coordinates, and then the Jacobian matrix can be derived by

$$J = \frac{\partial x}{\partial q} = \frac{\partial x}{\partial \varsigma} \cdot \begin{bmatrix} S(\xi) & 0_{7 \times n} \\ 0_{n \times 2} & I_{n \times n} \end{bmatrix} \quad (20)$$

2.3 Dynamic modelling

With the assumption of moving on a horizontal plane, the mobile platform has a constant potential energy U_m . To calculate the kinetic energy, the mobile platform can be divided into four parts, the cart (including all the driving units in the box), the left front wheel, the right front wheel and the rear wheel, that is

$$T_m = T_c + T_{lf} + T_{rf} + T_r \quad (21)$$

where T_m is the entire kinetic energy for the mobile platform; T_c , T_{lf} , T_{rf} and T_r denote respectively kinematic energy of the cart, the left and right front wheel and the rear wheel. Kinetic energy for the cart can be calculated by

$$T_c = \frac{1}{2} \cdot m_c \cdot \dot{x}_c^2 + \frac{1}{2} \cdot m_c \cdot \dot{y}_c^2 + \frac{1}{2} \cdot I_c \cdot \dot{\phi}_m^2 \quad (22)$$

where m_c is the mass of the cart; $x_c = x_m + l_c \cos \phi_m$, $y_c = y_m + l_c \sin \phi_m$ denote coordinates for the mass centre of the cart, with l_c being distance between the mass centre and O_m ; I_c represents the moment of inertia for the cart around the axis of $O_m Z_m$.

With the assumption of rolling without slipping, the contact points of the wheels with the motion plane can be treated as instant rotating centres. Then, kinematic energy for the front driving wheels can be given by

$$T_{lf} = \frac{3}{4} \cdot m_f \cdot r_f^2 \cdot \dot{\phi}_L^2, \quad T_{rf} = \frac{3}{4} \cdot m_f \cdot r_f^2 \cdot \dot{\phi}_R^2 \quad (23)$$

where m_f is the mass for the front driving wheels.

The castor wheel can rotate about both its axis and the fixed bar that does not pass through its centre, as shown in Fig. 3. So, the kinetic energy of the rear castor wheel includes two parts as given by Eq. 24.

$$T_r = \frac{1}{2} \cdot I_{r1} \cdot \dot{\phi}_r^2 + \frac{1}{2} \cdot I_{r2} \cdot \dot{\beta}_r^2 \quad (24)$$

where the moments of inertia I_{r1} and I_{r2} can be determined by

$$I_{r1} = \frac{1}{2} \cdot m_r \cdot r_r^2 + m_r \cdot r_r^2 = \frac{3}{2} \cdot m_r \cdot r_r^2 \quad (25)$$

$$I_{r2} = \frac{1}{4} \cdot m_r \cdot r_r^2 + m_r \cdot d_r^2 = m_r \cdot \left(\frac{r_r^2}{4} + d_r^2 \right)$$

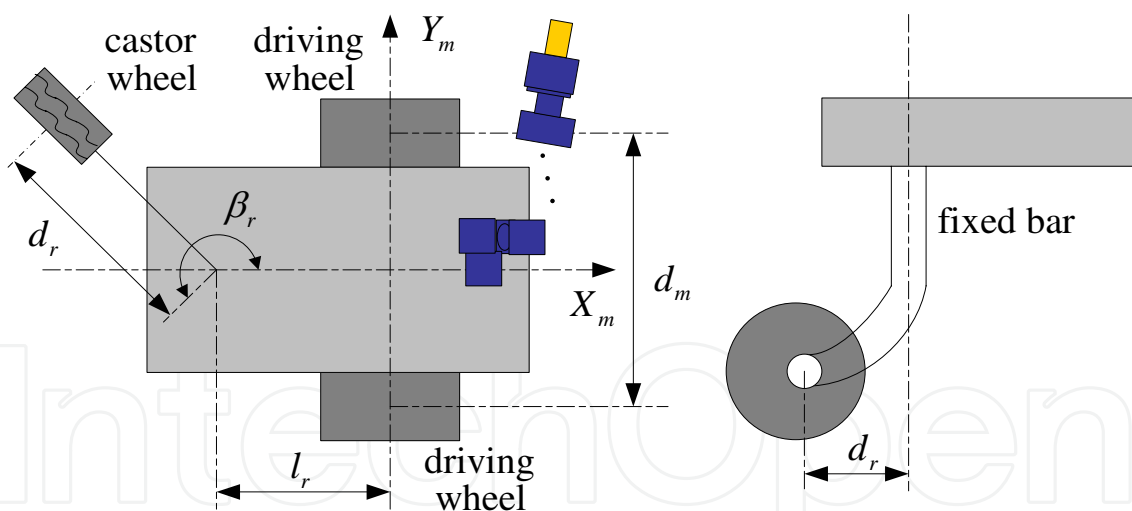


Fig. 3. Structure for the castor wheel.

Kinematic energy for the onboard modular manipulator can be calculated by

$$T_M = \sum_{i=0}^n \left(\frac{1}{2} \cdot v_{ci}^T \cdot m_i \cdot v_{ci} + \frac{1}{2} \cdot \omega_i^T \cdot I_i \cdot \omega_i \right) \quad (26)$$

where T_M denotes kinetic energy for the modular manipulator; m_i and I_i are mass and inertial moment matrix for the i^{th} module respectively; v_{ci} and ω_i represent the linear and angular velocities for the mass centre of the i^{th} module.

Potential energy for the onboard modular manipulator can be calculated by

$$U_M = \sum_{i=0}^n \{m_i \cdot g^T \cdot p_{ci}\} \quad (27)$$

where $g = [0 \ 0 \ 9.81]^T$ and $p_{ci} = [x_i^B \ y_i^B \ z_i^B]^T$.

Then, the Lagrange function can be given by

$$L = T_m - U_m + T_M - U_M \quad (28)$$

The constraint dynamics for the nonholonomic mobile modular manipulator can be determined by

$$\left\{ \begin{aligned} \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{x}_m} \right) - \frac{\partial L}{\partial x_m} &= \lambda_1 \cos \phi_m + \lambda_2 \cos \phi_m + \lambda_3 \sin \phi_m + \lambda_4 \cos(\beta_r + \phi_m) + \lambda_5 \sin(\beta_r + \phi_m) \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{y}_m} \right) - \frac{\partial L}{\partial y_m} &= \lambda_1 \sin \phi_m + \lambda_2 \sin \phi_m - \lambda_3 \cos \phi_m + \lambda_4 \sin(\beta_r + \phi_m) - \lambda_5 \cos(\beta_r + \phi_m) \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{\phi}_m} \right) - \frac{\partial L}{\partial \phi_m} &= -\lambda_1 \frac{d_m}{2} + \lambda_2 \frac{d_m}{2} - \lambda_4 l_r \sin(\beta_r) + \lambda_5 (l_r \cos \beta_r - d_r) \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{\phi}_r} \right) - \frac{\partial L}{\partial \phi_r} &= -\lambda_4 r_r \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{\beta}_r} \right) - \frac{\partial L}{\partial \beta_r} &= -\lambda_5 d_r \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{\phi}_L} \right) - \frac{\partial L}{\partial \phi_L} &= -\lambda_1 r_f + \tau_L + \frac{\partial x^T}{\partial \phi_L} F_{ext} \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{\phi}_R} \right) - \frac{\partial L}{\partial \phi_R} &= -\lambda_2 r_f + \tau_R + \frac{\partial x^T}{\partial \phi_R} F_{ext} \\ \frac{d}{dt} \left(\frac{\partial L}{\partial \dot{q}_i} \right) - \frac{\partial L}{\partial q_i} &= \tau_i + \frac{\partial x^T}{\partial q_i} F_{ext}, \quad (i=1, \dots, n) \end{aligned} \right. \quad (29)$$

where $\lambda_i (i=1, \dots, 5)$ are Lagrange multipliers associated with the nonholonomic constraints, and $F_{ext} \in \mathbb{R}^6$ is a vector for external forces; τ_L, τ_R, τ_i represent driving torques of the left wheel, the right wheel and the i^{th} joint for the onboard modular manipulator respectively.

Equation 29 can be written into the matrix form

$$\frac{d}{dt} \left(\frac{\partial L}{\partial \dot{\zeta}} \right)^T - \left(\frac{\partial L}{\partial \zeta} \right)^T = B(\zeta) \cdot (\tau + J^T \cdot F_{ext}) + C(\zeta) \cdot \lambda \quad (30)$$

where $B(\varsigma) = \begin{bmatrix} 0_{(n+2) \times 5} & I_{(n+2) \times (n+2)} \end{bmatrix}^T$, $C(\varsigma) = \begin{bmatrix} A(\xi) & 0_{5 \times n} \end{bmatrix}^T$, $\tau = [\tau_L \quad \tau_R \quad \tau_1 \quad \cdots \quad \tau_n]^T$, $\lambda = [\lambda_1 \quad \cdots \quad \lambda_5]^T$.

Equation 30 can be rewritten into the following standard form

$$M(\varsigma) \cdot \ddot{\varsigma} + V(\varsigma, \dot{\varsigma}) \cdot \dot{\varsigma} + G(\varsigma) = B(\varsigma) \cdot (\tau + J^T \cdot F_{ext}) + C(\varsigma) \cdot \lambda \quad (31)$$

where M is the inertial matrix; V represents the centrifugal and coriolis matrix; and G denotes the gravitational force vector.

Remark 1: Matrix M is symmetric and positive semi-definite, i.e. $M^T = M \geq 0$.

Remark 2: Matrix $\dot{M} - 2V$ is skew-symmetric, i.e., for any $r \in \mathbb{R}^{n+7}$, $r^T \cdot (\dot{M} - 2V) \cdot r = 0$.
From Eq. 20,

$$\dot{\varsigma} = \begin{bmatrix} S(\xi) & 0_{7 \times n} \\ 0_{n \times 2} & I_{n \times n} \end{bmatrix} \cdot \dot{q} \quad (32)$$

Substituting Eq. 32 and its derivative into Eq. 31 yields

$$\bar{M}(\varsigma) \cdot \ddot{q} + \bar{V}(\varsigma, \dot{q}) \cdot \dot{q} + \bar{G}(\varsigma) = \bar{\tau} \quad (33)$$

where $\bar{M} = \bar{S}^T \cdot M \cdot \bar{S}$, $\bar{V} = \bar{S}^T \cdot (V \cdot \bar{S} + M \cdot \dot{\bar{S}})$, $\bar{G} = \bar{S}^T \cdot G$, $\bar{\tau} = \tau + J^T \cdot F_{ext}$; the terms $\bar{S}^T \cdot B = I_{(n+2) \times (n+2)}$ and $\bar{S}^T \cdot C \cdot \lambda = 0_{n+2}$ are eliminated.

If $n < 4$, there are some dead zones that the end-effector can not reach; on the other hand, if $n > 4$ the entire mobile modular manipulator will be a redundant one; this two cases are beyond the discussion of this chapter. Assume the Jacobian matrix is full rank, solving the differential kinematics $\dot{x} = J \cdot \dot{q}$ and its derivative, yields

$$\dot{q} = J^{-1} \cdot \dot{x}, \quad \ddot{q} = J^{-1} \cdot \ddot{x} - J^{-1} \cdot \dot{J} \cdot J^{-1} \cdot \dot{x} \quad (34)$$

Substituting Eq. 34 into Eq. 33 and left multiplying J^{-T} yields

$$\tilde{M}(\varsigma) \cdot \ddot{x} + \tilde{V}(\varsigma, \dot{x}) \cdot \dot{x} + \tilde{G}(\varsigma) = \tilde{\tau} \quad (35)$$

where $\tilde{M} = J^{-T} \cdot \bar{M} \cdot J^{-1}$, $\tilde{V} = J^{-T} \cdot (\bar{V} - \bar{M} \cdot J^{-1} \cdot \dot{J}) \cdot J^{-1}$, $\tilde{G} = J^{-T} \cdot \bar{G}$ and $\tilde{\tau} = J^{-T} \cdot \bar{\tau}$.

Remark 3: Matrix \tilde{M} is symmetric and positive definite, i.e., $\tilde{M}^T = \tilde{M} > 0$.

Proof: From Eq. 33 and Eq. 35,

$$\tilde{M} = J^{-T} \cdot \bar{S}^T \cdot M \cdot \bar{S} \cdot J^{-1} \quad (36)$$

According to **Remark 1**,

$$\begin{aligned} \tilde{M}^T &= J^{-T} \cdot \bar{S}^T \cdot M^T \cdot \bar{S} \cdot J^{-1} = J^{-T} \cdot \bar{S}^T \cdot M \cdot \bar{S} \cdot J^{-1} = \tilde{M} \\ \tilde{M} &= J^{-T} \cdot \bar{S}^T \cdot M \cdot \bar{S} \cdot J^{-1} = (\bar{S} \cdot J^{-1})^T \cdot M \cdot (\bar{S} \cdot J^{-1}) > 0 \end{aligned} \quad (37)$$

Remark 4: Matrix $\dot{\tilde{M}} - 2\tilde{V}$ is skew-symmetric, i.e., for any $r \in \mathbb{R}^{n+7}$, $r^T \cdot (\dot{\tilde{M}} - 2\tilde{V}) \cdot r = 0$.

Proof: From Eq. 33 and Eq. 35,

$$J^T \cdot \tilde{M} \cdot J = \bar{S}^T \cdot M \cdot \bar{S} \quad (38)$$

Differentiating Eq. 38 yields

$$J^T \cdot \dot{\tilde{M}} \cdot J + 2 J^T \cdot \tilde{M} \cdot \dot{J} = \bar{S}^T \cdot \dot{M} \cdot \bar{S} + 2 \bar{S}^T \cdot M \cdot \dot{\bar{S}} \quad (39)$$

Then

$$\dot{\tilde{M}} = J^{-T} \cdot \bar{S}^T \cdot \dot{M} \cdot \bar{S} \cdot J^{-1} + 2 J^{-T} \cdot \bar{S}^T \cdot M \cdot \dot{\bar{S}} \cdot J^{-1} - 2 \tilde{M} \cdot \dot{J} \cdot J^{-1} \quad (40)$$

From Eq. 33 and Eq. 35

$$\tilde{V} = J^{-T} \cdot \bar{S}^T \cdot V \cdot \bar{S} \cdot J^{-1} + J^{-T} \cdot \bar{S}^T \cdot M \cdot \dot{\bar{S}} \cdot J^{-1} - \tilde{M} \cdot \dot{J} \cdot J^{-1} \quad (41)$$

Subtracting Eq. 41 multiplied by 2 from Eq. 40 yields

$$\dot{\tilde{M}} - 2\tilde{V} = J^{-T} \cdot \bar{S}^T \cdot (\dot{M} - 2V) \cdot \bar{S} \cdot J^{-1} \quad (42)$$

According to Eq. 41

$$r^T \cdot (\dot{\tilde{M}} - 2\tilde{V}) \cdot r = (\bar{S} \cdot J^{-1} \cdot r)^T \cdot (\dot{M} - 2V) \cdot (\bar{S} \cdot J^{-1} \cdot r) = 0 \quad (43)$$

Remark 5: Matrices \tilde{M} , \tilde{V} and \tilde{G} are all bounded as long as J keeps nonsingular.

Remark 6: The dynamic equation in Eq. 35 is linearity in parameters, i.e., for any $\chi \in \mathbb{R}^6$ independent of the structure parameters Φ , $\tilde{M}\ddot{\chi} + \tilde{V}\dot{\chi} + \tilde{G} = Y(\varsigma, \dot{q}, \ddot{\chi}, \dot{\chi})\Phi$. Here $Y(\varsigma, \dot{q}, \ddot{\chi}, \dot{\chi})$ is also independent of Φ .

3. Robust Adaptive Controller Design

3.1 A model-based controller

Assume x_d , \dot{x}_d and \ddot{x}_d are the desired end-effector positions, velocities and accelerations. If precise dynamic parameters can be obtained in advance, the model-based controller can be exploited, as shown in Fig. 4 and Eq. 44.

$$\tau = J^T \cdot \left\{ \tilde{M} \cdot [\ddot{x}_d + K_D \cdot (\dot{x}_d - \dot{x}) + K_P \cdot (x_d - x)] + \tilde{V} \cdot \dot{x} + \tilde{G} - F_{ext} \right\} \quad (44)$$

where $K_P > 0$ and $K_D^T = K_D > 0$ are proportional and differential gain matrices.

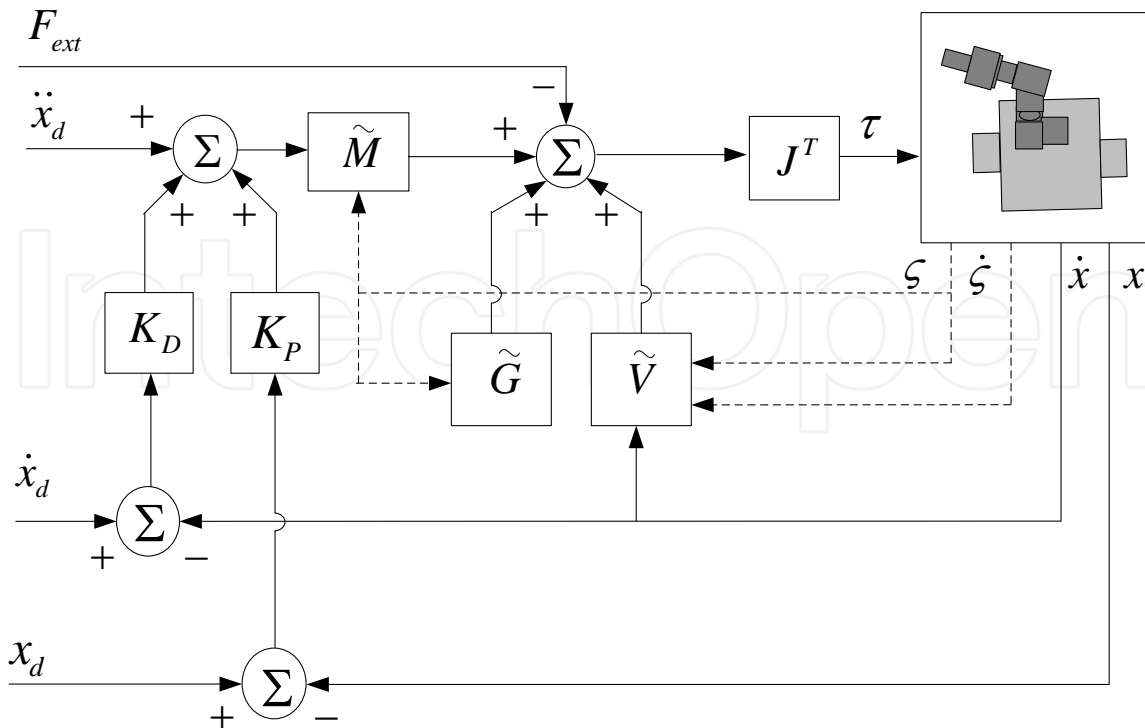


Fig. 4. A model-based controller.

Substituting Eq. 44 into Eq. 35 yields

$$\ddot{e} + K_D \cdot \dot{e} + K_P \cdot e = 0 \quad (45)$$

where $e = x - x_d$ is the error vector.

Theorem 1: If $K_P > 0$ and $K_D^T = K_D > 0$, the system described in Eq. 45 is asymptotically stable, i.e., $e \rightarrow 0$ as $t \rightarrow \infty$.

Proof: Considering the nonnegative Lyapunov candidate as follows

$$V_s = \frac{1}{2} \cdot \dot{e}^T \cdot \dot{e} + \frac{1}{2} \cdot e^T \cdot K_P \cdot e \geq 0 \quad (46)$$

The time derivative of Lyapunov candidate is

$$\dot{V}_s = \dot{e}^T \cdot \ddot{e} + \dot{e}^T \cdot K_P \cdot e \quad (47)$$

Substituting Eq. 45 into Eq. 47 yields

$$\dot{V}_s = -\dot{e}^T \cdot K_D \cdot \dot{e} \leq 0 \quad (48)$$

From Eq. 46 and Eq. 48, V_s is a Lyapunov function. Notice that when and only when $\dot{e} = 0$, $\dot{V}_s = 0$ may hold, so $\dot{e} \rightarrow 0$ as $t \rightarrow +\infty$. To analyze the steady-state error of the system, substituting $\dot{e} = 0$ and $\ddot{e} = 0$ into Eq. 45 yields $K_P \cdot e = 0$, which is followed by $e = 0$. That is to say, $e \rightarrow 0$ as $t \rightarrow +\infty$. That is all for the proof.

3.2 A robust adaptive controller

To avoid measuring accelerations, the error system can be defined as follows

$$\begin{aligned} e(t) &= x(t) - x_d(t) \\ \dot{x}_s(t) &= \dot{x}_d(t) - \Lambda \cdot e(t) \\ s(t) &= \dot{x}(t) - \dot{x}_s(t) \end{aligned} \quad (49)$$

where $s(t)$ is the tracking error measure, and $\Lambda \in \mathbb{R}^{6 \times 6}$ is a positive definite matrix.

From Eq. 49,

$$\begin{aligned} \dot{x}(t) &= s(t) + \dot{x}_s(t) \\ \ddot{x}_s(t) &= \ddot{x}_d(t) - \Lambda \cdot \dot{e}(t) \\ \ddot{x}(t) &= \dot{s}(t) + \ddot{x}_s(t) \end{aligned} \quad (50)$$

Substituting Eq. 50 into Eq. 35 yields

$$\tilde{M} \cdot \dot{s}(t) + \tilde{V} \cdot s(t) + \tilde{M} \cdot \ddot{x}_s(t) + \tilde{V} \cdot \dot{x}_s(t) + \tilde{G} = \tilde{\tau} \quad (51)$$

From **Remark 6**,

$$\tilde{M} \cdot \ddot{x}_s(t) + \tilde{V} \cdot \dot{x}_s(t) + \tilde{G} = Y(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \cdot \Phi \quad (52)$$

Substituting Eq. 52 into Eq. 51 yields

$$\tilde{M} \cdot \dot{s}(t) + \tilde{V} \cdot s(t) + Y(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \cdot \Phi = \tilde{\tau} \quad (53)$$

Let $\hat{\Phi}$ be the estimate of the structure parameter Φ , the robust adaptive controller presented in this chapter is given by Eq. 54, and a control system diagram is shown in Fig. 5.

$$\tau = J^T \cdot \left\{ Y(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \cdot \hat{\Phi} - K_p \cdot s(t) - K_I \cdot \int_0^t s(t) dt - \tau_r(s) - F_{ext} \right\} \quad (54)$$

where the 1st term forms an adaptive controller; the 2nd and 3rd terms form PI controller; $\tau_r(s)$ is the robust sliding mode term, with its element determined by

$$\tau_{ri}(s_i) = k_{ri} \cdot e^{\frac{1}{|s_i|}} \cdot \text{sgn}(s_i) \quad (55)$$

where $k_{ri} > 0$ is a constant.

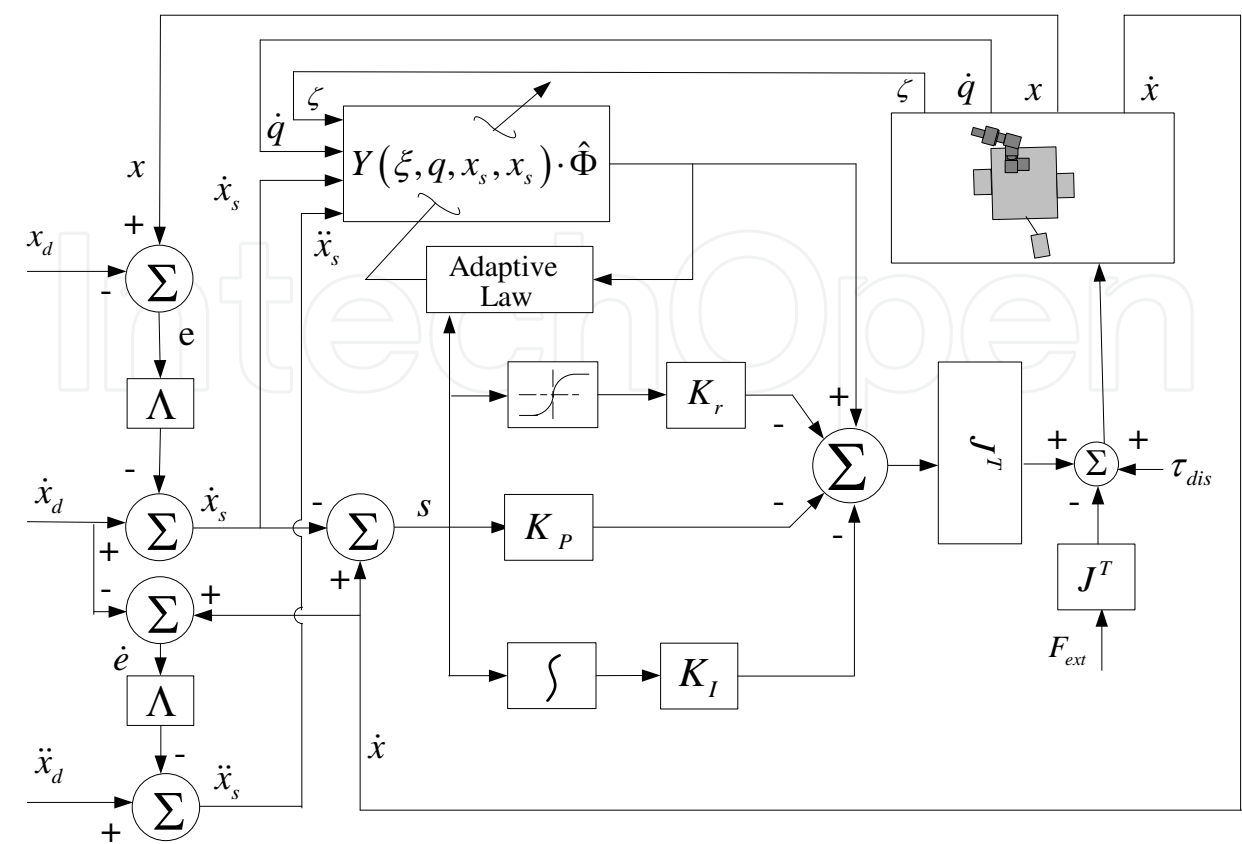


Fig. 5. A robust adaptive controller

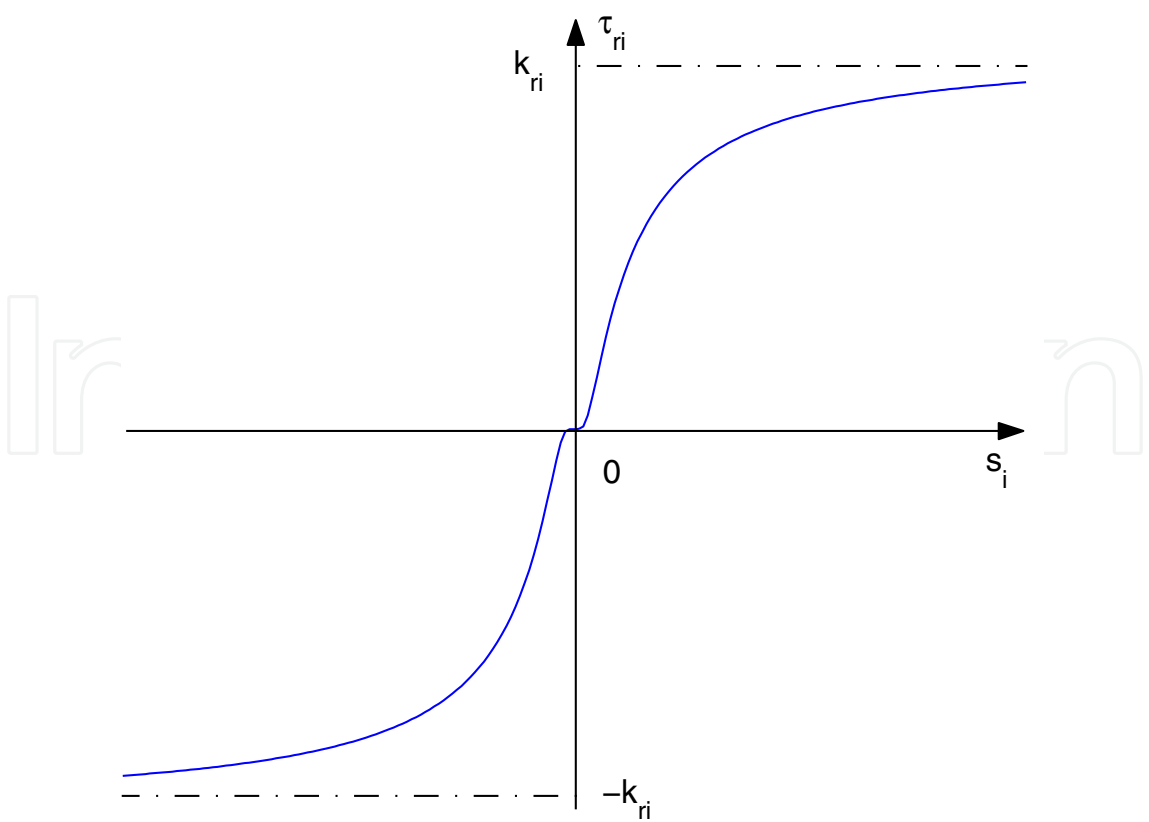


Fig. 6. A new sliding mode function.

The sliding mode control law is not only continuous but also infinitely differentiable, which can eliminate the chattering caused by the classical sign function effectively. The output varies according to $|s_i|$. The function is plotted in Fig. 6.

Substituting Eq. 54 into Eq. 53 and considering Eq. 35 at the same time yields

$$\tilde{M} \cdot \dot{s}(t) + \tilde{V} \cdot s(t) + K_p \cdot s(t) + K_I \cdot \int_0^t s(t) dt + Y(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \cdot (\Phi - \hat{\Phi}) + \tau_r = 0 \quad (56)$$

Theorem 2: The closed-loop system in Eq. 56 is asymptotically stable under the adaptation law by Eq. 57. The error signals are convergent with time, i.e., $e(t), \dot{e}(t) \rightarrow 0$ as $t \rightarrow \infty$. Furthermore, the signals in the system are all bounded.

$$\dot{\hat{\Phi}} = -\Gamma \cdot Y^T(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \cdot s \quad (57)$$

where $\Gamma^T = \Gamma > 0$ is a constant matrix.

Proof: Considering the nonnegative Lyapunov candidate as follows

$$V_s = \frac{1}{2} \cdot \left[\int_0^t s(t) dt \right]^T \cdot K_I \cdot \left[\int_0^t s(t) dt \right] + \frac{1}{2} \cdot s^T(t) \cdot \tilde{M} \cdot s(t) + \frac{1}{2} \cdot \tilde{\Phi}^T \cdot \Gamma^{-1} \cdot \tilde{\Phi} \quad (58)$$

where $\tilde{\Phi} = \hat{\Phi} - \Phi$.

Time derivative of the Lyapunov candidate is

$$\dot{V}_s = s^T \cdot \left[\tilde{M} \cdot \dot{s} + K_I \cdot \int_0^t s(t) dt \right] + \frac{1}{2} \cdot s^T \cdot \dot{\tilde{M}} \cdot s + \dot{\tilde{\Phi}}^T \cdot \Gamma^{-1} \cdot \tilde{\Phi} \quad (59)$$

From Eq. 56, we can obtain

$$s^T \left[\tilde{M} \cdot \dot{s} + K_I \cdot \int_0^t s(t) dt \right] = -s^T \cdot \tilde{V} \cdot s - s^T \cdot K_p \cdot s + s^T \cdot Y(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \tilde{\Phi} - s^T \tau_r \quad (60)$$

Substituting Eq. 60 into Eq. 59 and considering **Remark 4** at the same time yields

$$\dot{V}_s = -s^T \cdot K_p \cdot s + s^T \cdot Y(\zeta, \dot{q}, \ddot{x}_s, \dot{x}_s) \cdot \tilde{\Phi} - s^T \cdot \tau_r + \dot{\tilde{\Phi}}^T \cdot \Gamma^{-1} \cdot \tilde{\Phi} \quad (61)$$

Considering that $\dot{\tilde{\Phi}} = \dot{\hat{\Phi}}$, substituting Eq. 57 into Eq. 61 yields

$$\dot{V}_s = -s^T \cdot K_p \cdot s - s^T \cdot \tau_r \quad (62)$$

Substituting Eq. 55 into Eq. 62 yields

$$\dot{V}_s = -s^T \cdot K_p \cdot s - \sum_{i=1}^6 \left\{ k_{ri} \cdot e^{\frac{1}{|s_i|}} \cdot s_i \cdot \text{sgn}(s_i) \right\} \leq 0 \quad (63)$$

According to LaSalle's theorem, we can conclude that the system is asymptotically stable and $s \rightarrow 0$ as $t \rightarrow +\infty$. Let ℓ_p is the p -norm space, from Eq. 58 and Eq. 63, $s(t) \in \ell_2$. From

Eq. 50, $e(t) \in \ell_2 \cap \ell_\infty$, $\dot{e}(t) \in \ell_2$ and $e(t) \rightarrow 0$ as $t \rightarrow +\infty$. Because $V_s \geq 0$ and $\dot{V}_s \leq 0$, $\tilde{\Phi} \in \ell_\infty$ can be obtained; from **Remark 5**, $\Phi \in \ell_\infty$, so $\hat{\Phi} \in \ell_\infty$; therefore from the error equation, we can conclude that $\dot{s}(t) \in \ell_\infty$. Since $s(t) \in \ell_2$ and $\dot{s}(t) \in \ell_\infty$, $s \rightarrow 0$ as $t \rightarrow +\infty$, which is followed by $\dot{e}(t) \rightarrow 0$. This is the end of the proof.

4. Simulation Results

The mobile modular manipulator exploited in this simulation is made up of a 4-DOF onboard modular manipulator and 3-wheeled nonholonomic mobile platform, as shown in Fig. 1(a). The modular manipulator consists of 4 rotation modules named *PowerCube* produced by the *AmtecGmbH* Corporation of Germany. The *Pioneer3-DXe* produced by *ActiveMedia* Corporation of USA is used as the mobile platform. According to the Denavit-Hartenberg notion, the simplified model of the mobile modular manipulator is drawn in Fig. 7.

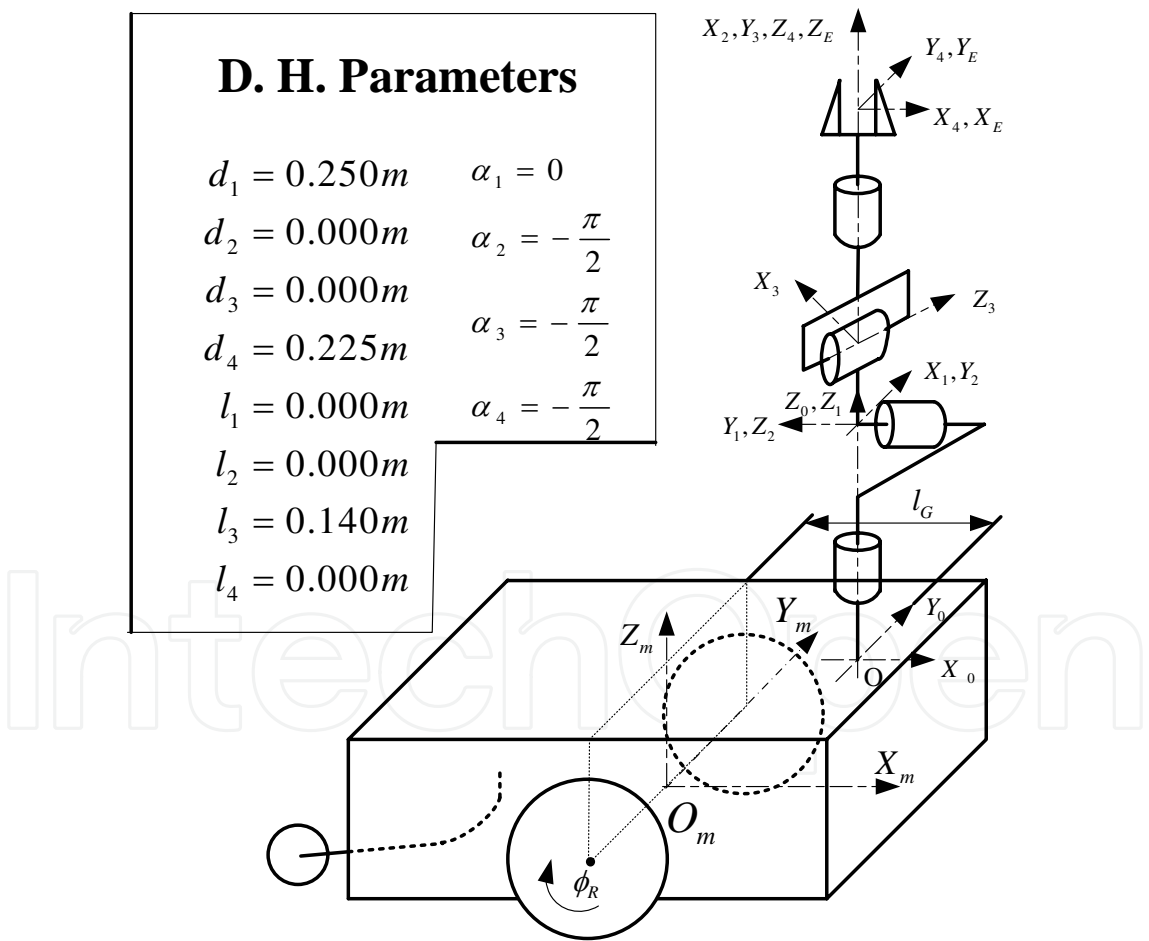


Fig. 7. A simplified model for the mobile modular manipulator.

Because of page limitations, the kinematics and dynamics model will not be detailed. To ensure the controllers valid, the desired end-effector trajectory should be selected as far away from singularities. The Jacobian matrix is derived from Eq. 20 as follows

$$J = \begin{bmatrix} J_{11} & J_{12} & J_{13} & J_{14} & J_{15} & 0 \\ J_{21} & J_{22} & J_{23} & J_{24} & J_{25} & 0 \\ 0 & 0 & 0 & -l_3 \cos q_2 + d_4 \cos q_2 \sin q_3 & d_4 \sin q_2 \cos q_3 & 0 \\ -\frac{r_f}{d_m} & \frac{r_f}{d_m} & 1 & \frac{\sin q_2 \sin q_3 \cos q_3}{\cos^2 q_3 + \cos^2 q_2 \sin^2 q_3} & \frac{-\cos q_2}{\cos^2 q_3 + \cos^2 q_2 \sin^2 q_3} & 0 \\ 0 & 0 & 0 & \frac{-\cos q_2 \sin q_3}{\sqrt{\cos^2 q_3 + \cos^2 q_2 \sin^2 q_3}} & \frac{-\sin q_2 \cos q_3}{\sqrt{\cos^2 q_3 + \cos^2 q_2 \sin^2 q_3}} & 0 \\ 0 & 0 & 0 & \frac{-\cos q_3}{\cos^2 q_2 + \sin^2 q_2 \cos^2 q_3} & \frac{\sin q_2 \cos q_2 \sin q_3}{\cos^2 q_2 + \sin^2 q_2 \cos^2 q_3} & 1 \end{bmatrix} \quad (64)$$

where

$$\begin{aligned} J_{11} &= \frac{r_f}{2} \cdot \cos \phi_m - \frac{r_f}{d_m} \cdot (J_{13} - l_G \cdot \sin \phi_m), & J_{21} &= \frac{r_f}{2} \cdot \sin \phi_m - \frac{r_f}{d_m} \cdot (J_{23} + l_G \cdot \cos \phi_m), \\ J_{12} &= \frac{r_f}{2} \cdot \cos \phi_m + \frac{r_f}{d_m} \cdot (J_{13} - l_G \cdot \sin \phi_m), & J_{22} &= \frac{r_f}{2} \cdot \sin \phi_m + \frac{r_f}{d_m} \cdot (J_{23} + l_G \cdot \cos \phi_m), \\ J_{13} &= -l_3 \sin(q_1 + \phi_m) \cos q_2 + d_4 \sin(q_1 + \phi_m) \cos q_2 \sin q_3 + d_4 \cos(q_1 + \phi_m) \cos q_3, \\ J_{23} &= +l_3 \cos(q_1 + \phi_m) \cos q_2 - d_4 \cos(q_1 + \phi_m) \cos q_2 \sin q_3 + d_4 \sin(q_1 + \phi_m) \cos q_3, \\ J_{14} &= -l_3 \cdot \cos(q_1 + \phi_m) \cdot \sin q_2 + d_4 \cdot \cos(q_1 + \phi_m) \cdot \sin q_2 \cdot \sin q_3, \\ J_{24} &= -l_3 \cdot \sin(q_1 + \phi_m) \cdot \sin q_2 + d_4 \cdot \sin(q_1 + \phi_m) \cdot \sin q_2 \cdot \sin q_3, \\ J_{15} &= -d_4 \cdot \cos(q_1 + \phi_m) \cdot \cos q_2 \cdot \cos q_3 - d_4 \cdot \sin(q_1 + \phi_m) \cdot \sin q_3, \\ J_{25} &= -d_4 \cdot \sin(q_1 + \phi_m) \cdot \cos q_2 \cdot \cos q_3 + d_4 \cdot \cos(q_1 + \phi_m) \cdot \sin q_3. \end{aligned} \quad (65)$$

The determinant of Jacobian matrix is

$$\det(J) = \frac{r_f^2 \cdot l_G \cdot l_3 \cdot \sin q_2 \cdot \cos q_2 \cdot \cos q_3}{d_m \cdot \sqrt{\cos^2 q_3 + \cos^2 q_2 \sin^2 q_3}} \quad (66)$$

So, to avoid singularities, the boundary of joint angles in the joint space can be selected as follows

$$\Omega = \left\{ \phi_L, \phi_R, q_1, q_2, q_3, q_4 \left| 0 < q_2 < \frac{\pi}{2}, -\frac{\pi}{2} < q_3 < \frac{\pi}{2} \right. \right\} \quad (67)$$

Simulations are conducted for two control schemes, i.e., the model-based controller (MBC) and the robust adaptive controller (RAC). To examine the disturbance suppression characteristics, a series of disturbance torques are introduced. Both the MBC and the RAC are required to control the end-effector to follow a sine like spatial trajectory as shown in Fig. 8, which has been planned to ensure the system far away from singularities. All the joints and velocities are initialized to be zeros, except that $q_2 = \pi/3$. Simulation time interval is selected as 10 seconds. The gain matrices for the RAC are selected as follows

$$\begin{aligned} K_p &= \text{diag}\{100\}, \quad K_I = \text{diag}\{10.0\}, \quad K_r = \text{diag}\{10.0\}, \\ \Lambda &= \text{diag}\{2.0\}, \quad \Gamma = \text{diag}\{0.1\}. \end{aligned} \quad (68)$$

For the MBC, nominal dynamic parameters are adopted, which are assumed to be deviated from the real values by 10%; the gain matrices are selected as $K_p = \text{diag}\{100\}$, $K_D = \text{diag}\{10.0\}$.

Simulation results are presented by Fig. 8-14. The desired and controlled locus for the MBC and the RAC are shown in Fig. 8(a) and Fig. 8(b); Figure 9(a) and Fig. 9(b) present the tracking position errors; tracking Euler angular errors are shown in Fig. 10(a) and Fig. 10(b); Figure 11(a) and Fig. 11(b) give the tracking linear velocity errors; tracking Euler angular velocity errors are given by Fig. 12(a) and Fig. 12(b); time-varying control torques for the MBC and the RAC are shown in Fig. 13 and Fig. 14.

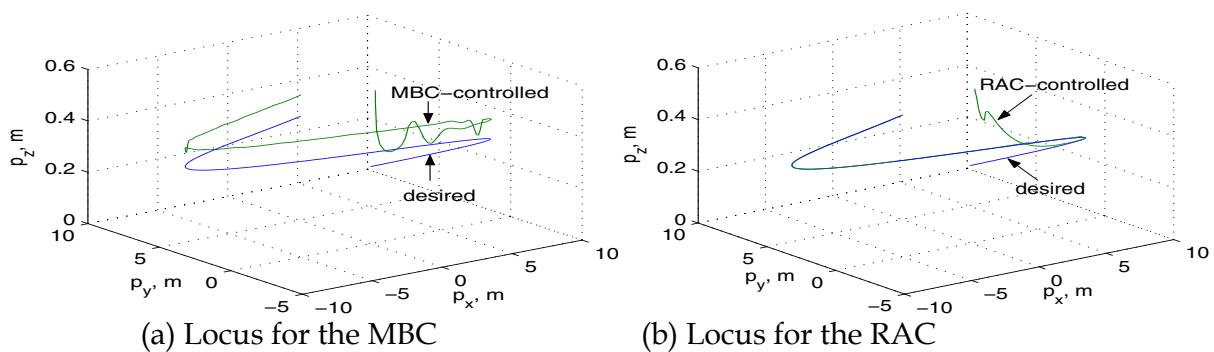


Fig. 8. Desired and controlled locus for the MBC and RAC.

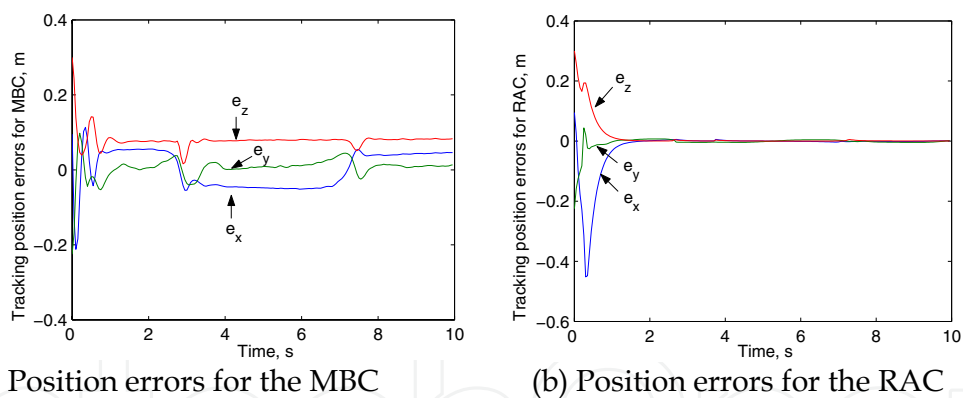


Fig. 9. Tracking position errors for the MBC and RAC.

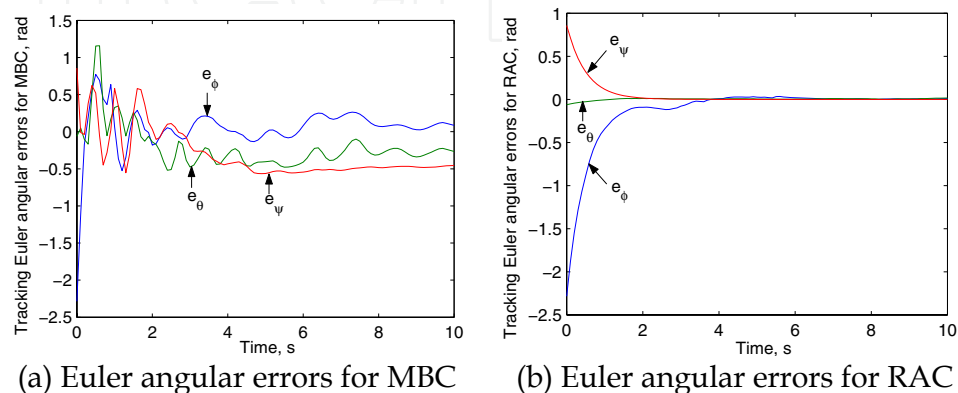


Fig. 10. Tracking Euler angular errors for the MBC and RAC.

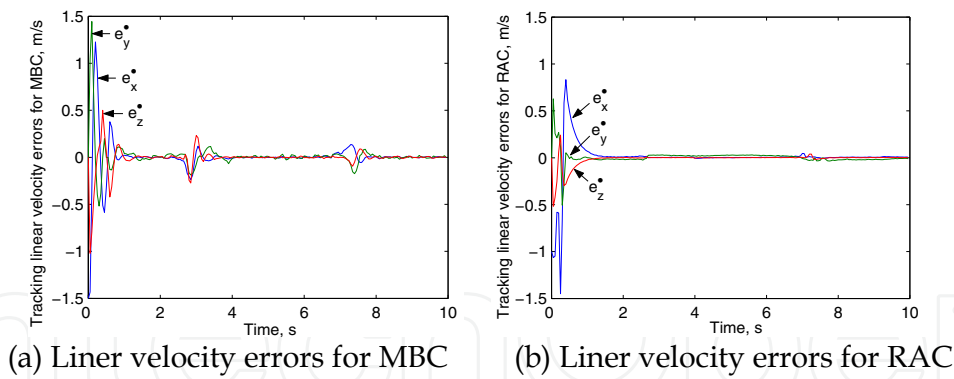


Fig. 11. Tracking linear velocity errors for the MBC and RAC.

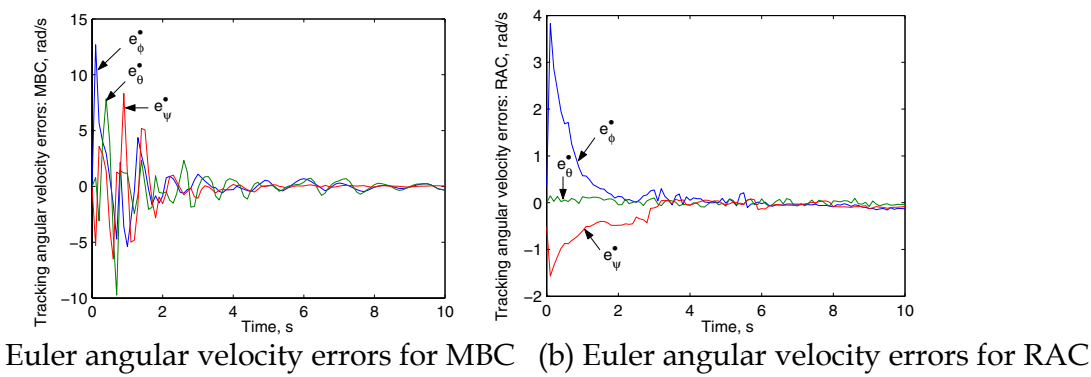


Fig. 12. Tracking Euler angular velocity errors for the MBC and RAC.

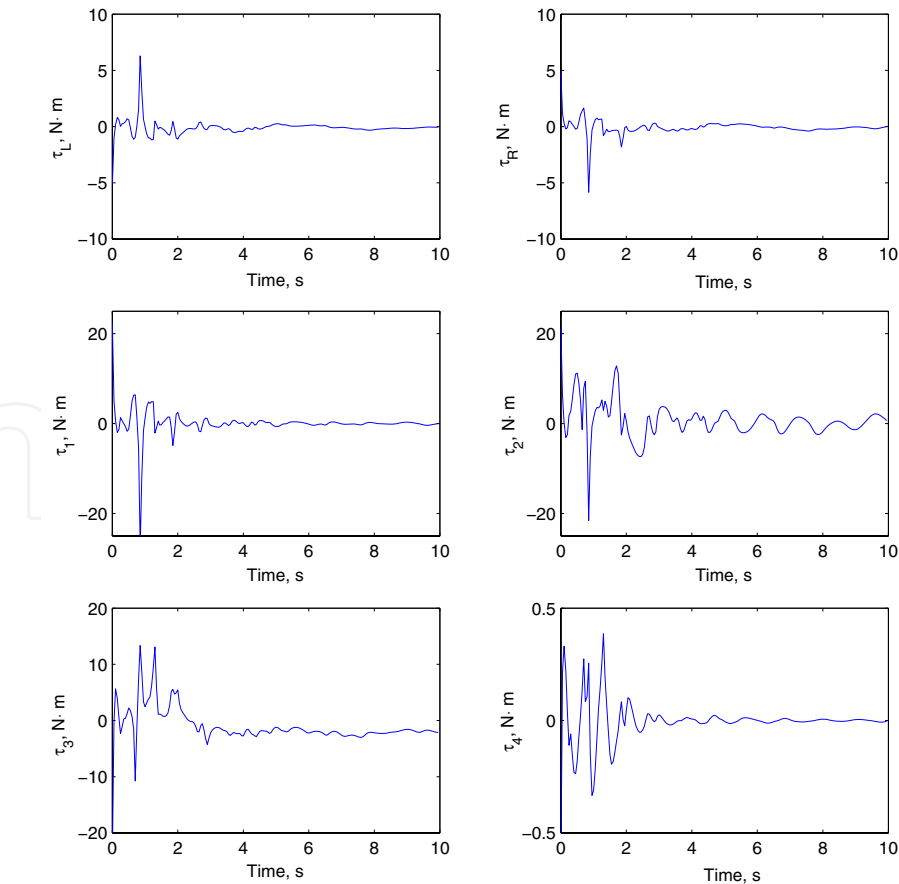


Fig. 13. Time-varying control torques for the MBC.

Remark 7: From the simulation results, we can conclude that the proposed algorithms are effective to control the end-effector to follow some definite spatial trajectories. The robust adaptive controller behaves better than the model based controller, especially when parameter uncertainties and external disturbances exist.

5. Conclusions

In this chapter, 3-wheeled nonholonomic mobile modular manipulators are investigated. First, an integrated modelling method is proposed in consideration of the nonholonomic constraints and the interactive motions. Second, a model based controller and a robust adaptive controller are presented in task space directly. Third, simulations are carried out on a mobile modular manipulator composed of a 3-wheeled mobile platform and a 4-DOF onboard modular manipulator, and the simulation results demonstrate the effectiveness of the proposed algorithms. The proposed algorithms can be easily extended to some other mobile manipulators as well.

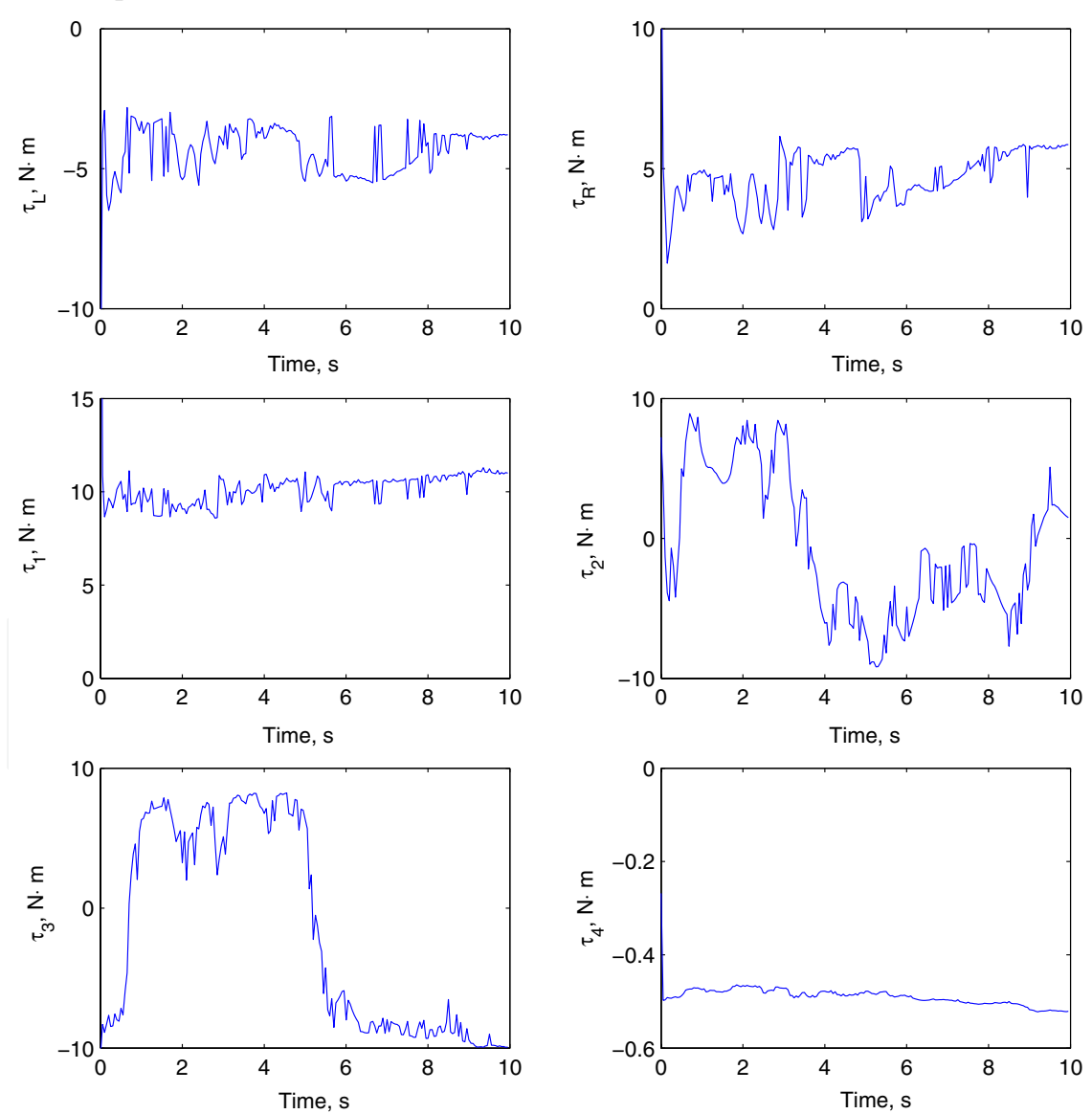


Fig. 14. Time-varying control torques for the RAC.

6. Future Research Work

In this chapter, although the number of joints for the onboard modular manipulator is just assumed to be 4, the proposed method can be applied to m-DOF manipulator atop n-DOF mobile platform system in principle. The modelling and control for a redundant mobile modular manipulator is more challenging since self-motions have to be taken into consideration, (Li & Liu, 2006b; Maciejewski & Klein, 1989). The proposed algorithms are just verified by simulations, there is still a lot of work to do to verify the algorithms by real experiments. The task space information in this simulation is calculated via forward kinematics and differential kinematics, how to acquire task-space information directly and precisely is also a complex work, which needs deep research work on sensory technology and method.

7. Acknowledgement

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This book covers many aspects of the exciting research in mobile robotics. It deals with different aspects of the control problem, especially also under uncertainty and faults. Mechanical design issues are discussed along with new sensor and actuator concepts. Games like soccer are a good example which comprise many of the aforementioned challenges in a single comprehensive and in the same time entertaining framework. Thus, the book comprises contributions dealing with aspects of the Robotcup competition. The reader will get a feel how the problems cover virtually all engineering disciplines ranging from theoretical research to very application specific work. In addition interesting problems for physics and mathematics arises out of such research. We hope this book will be an inspiring source of knowledge and ideas, stimulating further research in this exciting field. The promises and possible benefits of such efforts are manifold, they range from new transportation systems, intelligent cars to flexible assistants in factories and construction sites, over service robot which assist and support us in daily live, all the way to the possibility for efficient help for impaired and advances in prosthetics.

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