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An Exact Impedance Control of DC Motors Using Casimir Function

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1. Introduction

This chapter gives a new and exact impedance control of DC motor. From Hogan's original work, the control input for impedance control is torque input since the impedance control is designed for Lagrangian systems. However, in actual situation, there exist dynamics between the torque and control input and this dynamics can be dominant in certain scale. In such situation, if we neglect the dynamics or try to cancel the dynamics, the standard impedance control can lose the stability or the control performance at least.

To overcome this problem, we need an new impedance control which takes the dynamics into account without canceling any dynamics. In this chapter we give a solution for this problem by focusing on Casimir function which is rarely used in the conventional robotics.

The rest of this chapter is organized as follwows. In Section 2, we give a new model of DC motor with dynamics between the torque and control input. In Section 3, we propose a new impedance control which is based on Casimir function. Casimir function is one of the properties of port-Hamiltonian systems. In Section 4, we confirm the proposed method in numerical simulation and we conclude this chapter in Section 5.

2. Modeling

Let us start from a well-known model of DC motor:

$$\begin{bmatrix} \dot{\theta} \\ \dot{\omega} \\ \dot{i} \end{bmatrix} = \begin{bmatrix} 0 & \frac{1}{J} & 0 \\ 0 & 0 & K \\ 0 & -\frac{K}{T} & -\frac{R}{T} \end{bmatrix} \begin{bmatrix} \theta \\ \omega \\ i \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \frac{1}{J} \end{bmatrix} v \tag{1}$$

where the displacement θ , the velocity ω and the current i are the states , the voltage v is the control input with the torque constant K and the inductance L.

Although the system (1) is a third-order system and thus not mechanical system, the system (1) has a mechanical-like structure, that is, can be modeled as a port-Hamiltonian system van der Schaft (2000), Maschke & van der Schaft (1996) with a Hamiltonian $H = (1/2J)p^2 + (1/2)Kr^2$

$$\begin{bmatrix} \dot{q} \\ \dot{p} \\ \dot{r} \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 & 0 \\ -1 & 0 & \bar{K} \\ 0 & -\bar{K} & -\bar{R} \end{bmatrix}}_{A} \nabla_{x} H + \begin{bmatrix} 0 \\ 0 \\ \bar{v} \end{bmatrix}$$
(2)

where $x = \begin{bmatrix} q & p & r \end{bmatrix}^T = \begin{bmatrix} \theta & J\omega & \sqrt{L}i \end{bmatrix}^T$ is the new state, $\bar{v} = v/\sqrt{L}$ is the new input with $\bar{K} = \frac{K}{\sqrt{L}} \bar{R} = \frac{R}{L}$ and

$$y = \nabla_r H$$

is taken as the passive output. It is confirmed that this system is now passive Takegaki & Arimoto (1981) with respect to the Hamiltonian H, that is,

$$\dot{H} \leq y^{\mathrm{T}} v$$

holds. This passive property is studied in many systems and used as a structural property to drive robust and nonliear controllers for stabilization Ortega & Garcia-Canseco (2004) Stramigioli et al. (1998), trajectory tracking Fujimoto & Sugie (2001) and motion generation problems Sakai & Stramigioli (2007).

However, in this chapter, we consider a different problem, namely, impedance control problems and we do not focus on the passivity but focus on another structural property:

Lemma 1 Consider the system (1) with zero-input $v \equiv 0$ in the case of no dissipation R = 0. Let the skew-symmetric part of the matrix A be $J(x) = J(x)^T$. Then the system has a solluton of the following PDE

$$\nabla_x C(x)J(x) = 0$$

and the solution is characterized as

$$C(x) = \bar{K}q + r.$$

Proof This is confrimed by a direct calculation. (Q.E.D.)

This means that, in the case of no dissipation R = 0, not only the Hamiltonian function H but also the Casimir function C are constant

$$\dot{C} = 0 \ (u \equiv 0)$$

for any the value of the Hamiltonian function H. Then we can express the system (1) by using the Casimir function (with respect to J) explicity.

Lemma 2 (Modeling) Consider the system (1) with zero-input $v \equiv 0$ in the case of no dissipation R = 0. Then the coordinate transformation convert the system (1) into

$$\begin{bmatrix} \dot{q} \\ \dot{p} \\ \dot{C} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ -1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \nabla_{x_c} \bar{H} + \begin{bmatrix} 0 \\ 0 \\ \bar{v} \end{bmatrix}$$
(3)

with the state $x_c = (q \ p \ C)^T$ and the Hamiltonian function

$$\bar{H} = \frac{p^2}{2I} + \frac{\left(\bar{K}q\right)^2}{2} - \bar{K}qC.$$

Proof This is also confrimed by a direct calculation although the old Hamiltonian H is not equal to the new Hamiltonian \bar{H} . (Q.E.D.)

3. Exact impedance control

In this section, we give an exact impedance control for DC motor by using Casimir functions. **Proposition 1 (Main result)** *Consider the system (1) with the velocity input. Then the following controller*

$$\begin{cases}
\dot{\bar{C}} = \bar{K}q - (1 + k_c)(\bar{K}q + r) \\
v = \bar{C}/J_c
\end{cases} \tag{4}$$

converts the close-loop system into the mechanical system with the impedance parameters J_c , $k_c > 0$. **Proof** First we introduce an artificial Casimir function \bar{C} and via the following dynamic extension

where \bar{v} is the input corresponding to the (artificial) Casimir function.

Then the Hamiltonian function \bar{H} is replaced by the following new Hamiltonian function which has a special structure suitable for impedance design with any parameters $k_c > 0$ and $J_c > 0$ as follows:

$$\bar{H}_{mec} = \bar{H} + \bar{K}\frac{C^2}{2} + \frac{\bar{C}^2}{2J_c} + K_c \frac{C^2}{2}$$

$$= \frac{p^2}{2J} + \frac{\bar{K}}{2} \left(q - \frac{C}{\bar{K}} \right)^2 + \frac{k_c C^2}{2} + \frac{\bar{C}^2}{2J_c} \tag{6}$$

due to the definition of the Casimir function. Finally the dynamic controller

$$\begin{cases}
v = +\nabla_{\bar{C}} H_{mec} \\
\bar{v} = -\nabla_{C} H_{mec}
\end{cases} \tag{7}$$

converts the system (1) with a dissipation $R \ge 0$ into the following Hamiltonian system

Fig. 1. A port-Hamiltonian system with flow inputs.

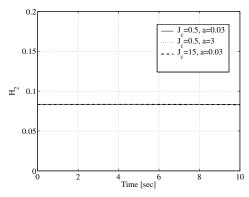


Fig. 2. Time response of H_{mec}

with
$$\bar{x}_c = (q \ p \ C \ \bar{C})^{\mathrm{T}}$$
. (Q.E.D.)

The proposed impedance control does not input the torque but the velocity, unlike the conventional impedance control for the mechanical systems. This difference is ullustrated in Fig.1. The spring coefficient k between the real mass and the virtual mass is not design parameter unlike the spring coefficient k_c between the environment and the virtual mass. Note that there is no canceling action in the controller.

4. Numerical simulations

Fig.2 shows the time response of the Hamiltonian function H_{mec} in the case of no dissipation (the Adams method) in the case of the parameters J = 1.5 L = 0.165 K = 0.47, $J_c = 0.5$, a = 0.03 and the initial conditions r(0) = q(0) = 0 p(0) = 0.5. It is confrimed that the value is constant as in the acutal Hamiltonian systems

Figs.3-5 show the time responses of the Casmir function and the state q and p in the case of dissipation R = 3.2. The parameters have changed $J_c \rightarrow 15$ and $a \rightarrow 3$.

In all cases, the nonliner behavior has changed intuitively due to the mechanial strucutre in the closed-loop system. The validity of our methods are confirmed.

5. Conclusions

there exist dynamics between the torque and control input and this dynamics can be dominant in certain scale. In such situation, if we neglect the dynamics or try to cancel the dynamics, the standard impedance control can lose the stability or the control performance at least.

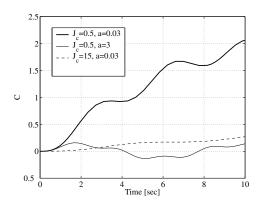


Fig. 3. Time response of C(t)

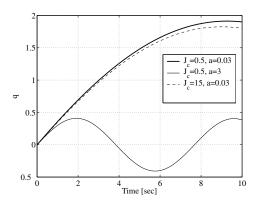


Fig. 4. Time response of q(t)

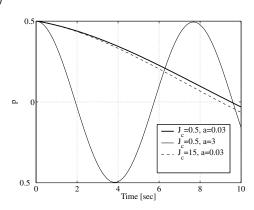


Fig. 5. Time response of p(t)

To overcome this problem, we need an new impedance control which takes the dynamics into account without canceling any dynamics. In this chapter we give a solution for this problem by focusing on Casimir function which is rarely used in the conventional robotics.

First we give a new model of DC motor with dynamics between the torque and control input. Second, we propose a new impedance control which is based on Casimir function. Casimir function is one of the properties of port-Hamiltonian systems. Finally, we confirm the proposed method in numerical simulation.

The generalization of the proposed method and appilications to other systems (such as hydraulic systems and mascle-skelton systems) are next works in near future.

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7. References

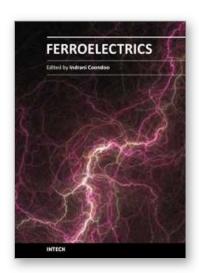
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Ferroelectric materials exhibit a wide spectrum of functional properties, including switchable polarization, piezoelectricity, high non-linear optical activity, pyroelectricity, and non-linear dielectric behaviour. These properties are crucial for application in electronic devices such as sensors, microactuators, infrared detectors, microwave phase filters and, non-volatile memories. This unique combination of properties of ferroelectric materials has attracted researchers and engineers for a long time. This book reviews a wide range of diverse topics related to the phenomenon of ferroelectricity (in the bulk as well as thin film form) and provides a forum for scientists, engineers, and students working in this field. The present book containing 24 chapters is a result of contributions of experts from international scientific community working in different aspects of ferroelectricity related to experimental and theoretical work aimed at the understanding of ferroelectricity and their utilization in devices. It provides an up-to-date insightful coverage to the recent advances in the synthesis, characterization, functional properties and potential device applications in specialized areas.

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